

Article

Joint Beamforming Design for RIS-assisted Integrated Satellite-HAP-Terrestrial Networks Using Deep Reinforcement Learning

Min Wu¹, Shibing Zhu¹, Changqing Li¹, Yudi Chen^{1,*}, Feng Zhou² and Xiang Su³

¹ Space Engineering University; 1800022837@pku.edu.cn (M. W.); sbz_zhu@sohu.com (S. Z.); lcqqcl5577@sohu.com (C. L.); cheniyudi9438@163.com (Y. C.);

² Yancheng Institute of Technology, Yancheng, China; zfyct@ycit.edu.cn (F. Z.);

³ 714 Research Institute of China State, Beijing, China; sx_fly@0603@163.com (X. S.);

* Correspondence: sbz_zhu@sohu.com;

Abstract: In this paper, we consider a reconfigurable intelligent surface (RIS)-assisted integrated satellite-high altitude platform-terrestrial networks (IS-HAP-TNs) that can improve network performance by exploiting HAP's stability and RIS's reflection. Specifically, the reflector RIS is installed on the side of HAP to reflect signals from the multiple ground user equipments (UEs) to the satellite. To aim at maximising system sum rate, we jointly optimize the transmit beamforming matrix at the ground UEs and RIS phase shift matrix. Due to the limitation of the unit modulus of the RIS reflective elements constraint, the combinatorial optimization problem is difficult to tackle it effectively by traditional solving methods. Based on this, this paper studies deep reinforcement learning (DRL) algorithm to achieve online decision making for this joint optimization problem. In addition, it is verified through simulation experiments that the proposed DRL algorithm outperforms the standard scheme in terms of system performance and execution time, and higher computing speed, making real-time decision making truly feasible.

Keywords: Reconfigurable intelligent surface (RIS); integrated satellite-HAP-terrestrial networks (IS-HAP-TNs); deep reinforcement learning (DRL); optimization performance;

1. Introduction

As fifth-generation mobile communication systems enter commercial operation worldwide, terrestrial wired and wireless networks are beginning to provide instant, high-speed data transmission services to users in high-density population areas, but due to geographical conditions and business models, networks in remote areas are still unable to meet multiple users' needs for full-area coverage and ubiquitous access. Compared with traditional terrestrial wireless communication systems, the integration of satellite, aerial platform and terrestrial communications into the integrated satellite-high altitude platform-terrestrial networks (IS-HAP-TNs) have emerged as a very potential infrastructure in the future wireless communication networks, which can establish seamless coverage and massive connectivity for the explosive growth of terrestrial users [1,2]. Nevertheless, IS-HAP-TNs also raise serious concern about the rapidly growing energy consumption and wireless security in the transmission process, which are of great significance for maintaining green and reliable communication schemes [3].

Among the various candidates, a novel energy-efficient mode, known as reconfigurable intelligent surface (RIS), has been widely applied to improve communication security and network performance[4,5]. Each of the RIS reflective element is a varactor diode that allows the amplitude and/or phase shift of the incident signal to be independently controlled by an embedded RIS central controller [6]. An extensive study in [7] shows that RIS has already been applied in many different communication network scenarios, such as ambient reflectors, signal transmitters, and even signal receivers.

36 Meanwhile, RIS is also used in ambient forward scatter/backscatter communication
37 systems, which is a seminal contribution, as in [8].

38 Recently, deep reinforcement learning (DRL) has made a splash in non-convex
39 optimization problems, including hybrid beamforming design[14], spectrum intelligence
40 sensing [15], channel state estimation [16], and power allocation strategy optimization
41 [17]. Compared with deep learning (DL), the DRL algorithm does not require a large
42 amount of training labeled data as inputs and is therefore very friendly for optimization
43 of wireless communication systems where obtaining data is more tedious. By inter-
44 acting with the environment to obtain rewards from the network, DRL can learn and
45 construct wireless channel knowledge without knowing the complete channel model
46 information and the precise movement pattern, while implementing efficient algorithm
47 design through embedded neural networks to sequentially find optimal solutions to
48 complex multi-objective optimization problems. In [18], a deep Q-network (DQN) with
49 greedy characteristics is proposed for the jointly optimizing of beamforming design,
50 power allocation strategy and interference coordination for maximizing the signal to
51 interference plus noise ratio (SINR). In [19], by using the DRL framework, the user
52 distribution model is tracked and predicted to autonomously and dynamically optimize
53 the MIMO broadcast beam and propose the optimal broadcast beam for each served cell.
54 The results confirm that optimal coverage can be achieved using the DRL framework in
55 both single-sector and multi-sector environments, and in both periodic and Markovian
56 mobility modes.

57 Currently, the joint beamforming design technique is also a hot issue, which can
58 greatly improve the communication efficiency, system capacity and transmission rate of
59 wireless communication systems to some extent. Motivated by the above analysis, our
60 objective is to design a novel DRL framework that considered as the DDPG algorithm
61 for the jointly optimization problem under the proposed RIS-assisted IS-HAP-TNs [20].
62 In particular, considering the high dynamics of RIS-assisted IS-HAP-TNs transmission
63 process, the main work and contributions of this paper can be summarized as follows

- 64 • Firstly, considering the time-varying characteristics of IS-HAP-TNs fading channel
65 model and signal transmission model, the system sum rate formulations are given
66 under this system model constraints using the active transmit beamforming at the
67 ground user equipments and the phase shift matrix at the RIS, and the maximization
68 expressions under the proposed constraints.
- 69 • Secondly, a soft-update strategy framework based on DDPG framework is designed
70 to optimize the above target problems. The framework does not need to know the
71 explicit model and specific mobile model of wireless environment, and can well
72 deal with the continuous state space, action space and reward function, and solve
73 the formal problem of the system.
- 74 • Finally, the simulation experiments on the number of RIS elements as well as the
75 average reward show that the designed DRL algorithm framework outperforms
76 other algorithms, which is a guideline for real-time decision making in dynamic
77 IS-HAP-TNs communication environments.

78 The remainder of this paper is arranged as follows. Section II describes the con-
79 sidered system model and identifies the optimization objective problem under the
80 constraints. Section III gives the basic framework of the soft update parameter strategy
81 and gives the design flow for the optimization of the active transmit beamforming ma-
82 trix and the RIS phase shift matrix under this framework. Section IV plots the network
83 performance simulation results under this framework and provides a detailed theoretical
84 analysis. Finally, Section V concludes the whole work.

85 2. System Model Description

86 In this illustration, we envision an uplink transmission communication system
87 includes geosynchronous earth orbit (GEO) satellite, backward high altitude platforms
88 (HAPs) deployed with RIS, the K ground user equipments (UEs) employs a single

89 antenna as shown in Fig. 1. In our proposed system model, the UEs transmission
 90 communication information through RF links to the RIS which installed on the HAP
 91 with M reflective elements, which acts as a reflecting relay with changeable transmission
 92 links, and sends the received signal to the satellite.

93 It is noted that the satellite are linked to the cloud data computing processing center
 94 by free-space optical (FSO), which can collect global communication information such as
 95 the user's requirements as system control link. Instead of coding satellites and HAPs
 96 separately, it centralizes the baseband processing of the entire network in the cloud,
 97 with the cloud as the core, taking into account resource management and environmental
 98 feedback [21].

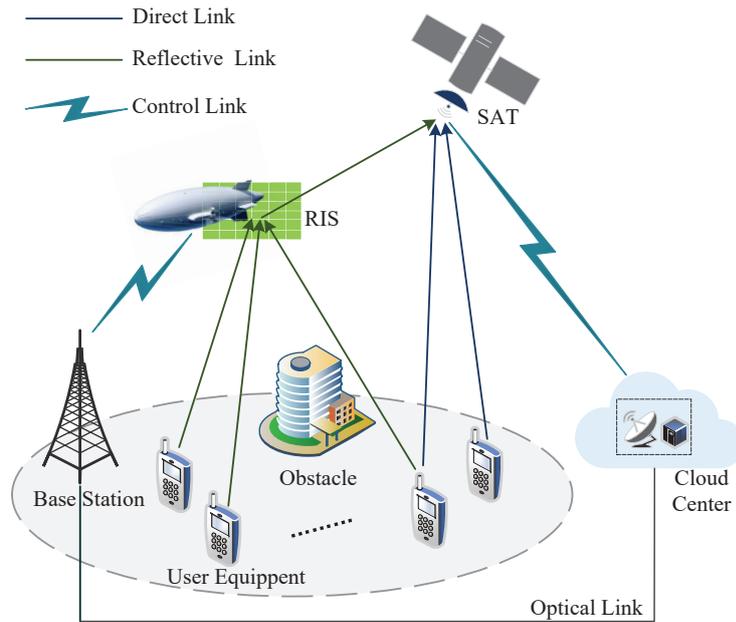


Figure 1. Illustration of a RIS-assisted IS-HAP-TNs system

In order to realistically simulate the UEs-RIS link, where the RIS is mounted on the HAP in the aerial. Here, we consider the small-scale path loss model [22], then the channel vector of the UEs-RIS can be expressed as

$$\mathbf{h}_{UR} = \sqrt{\frac{MK}{L_{total}}} \sum_{l=1}^{L_{total}} \alpha_l \mathbf{g}(m, \varphi_{AR}) \mathbf{g}^T(k, \varphi_{DU}) \quad (1)$$

where L_{total} denotes the number of the total transmission path, α_l represents the Nakagami- m channel model random variable, $\varphi_{AR,l}$ and $\varphi_{DU,l}$ denote the the angle of arrival (AoA) of RIS and the angle of departure (AoD) of the UEs in the l -th transmission path. The channel model vector $\mathbf{g}(L, \varphi)$ as a function of the transmission path L and the AOA or AOD φ can be expressed as

$$\mathbf{g}(L, \varphi) \triangleq \frac{1}{\sqrt{L}} [1, e^{j\pi \cos \varphi}, e^{2j\pi \cos \varphi}, \dots, e^{(L-1)j\pi \cos \varphi}]^T \quad (2)$$

The RIS-satellite uplink channel vector is denoted by \mathbf{H}_{RS} , which can expressed as

$$\mathbf{H}_{RS} = \sqrt{MN_s P_r} [\mathbf{g}(N_s, \varphi_{AS})] \mathbf{g}^T(M^T, \varphi_{DR}) \quad (3)$$

where the N_s denotes the antenna numbers of the uniform linear array (ULA) in the satellite, φ_{AS} and φ_{DR} are the AOA of the satellite and the AOD of the RIS, respectively. Meanwhile, the P_r is the free space path loss between the RIS and the satellite [23]. Note

that in the RIS-satellite uplink channel model, considering that the HAP flies at a higher altitude than most ground buildings and the RIS is mounted on the HAP, thus we only assume the line-of-sight (LoS) transmission path between the RIS and the satellite, the P_r can be expressed as by the following formula

$$P_r = \frac{\lambda^2 G_{sr} G_{st}}{(4\pi\tau)^2 d_{sr}^2 \kappa_a T_a B_W} \quad (4)$$

99 where λ , G_{sr} , G_{st} , d_{sr} , κ_a , T_a , B_W denote the carrier wavelength of signal, the gains
100 of every RIS reflection unit, antenna gain of each satellite, the transmission distance
101 between RIS to center of satellite coverage area, the Boltzmann constant, the temperature
102 of the propagating noise and the frequency band of signal, respectively. The direct uplink
103 channel \mathbf{H}_{US} from the UE and the satellite is basically a standard MIMO channel and
104 can be characterized by existing methods to express its channel characteristics [24].

We defined that Φ means diagonal matrix for input by $\Phi(m, m) = \phi_m = \chi_m e^{j\varphi_m}$, $m = 1, 2, \dots, M$, $\chi_m \in [0, 1]$, representing the magnitude and phase shift of each RIS element, respectively. It is supposed that \mathbf{H}_{RS} and \mathbf{H}_{US} remain unchanged in consecutive K consecutive time slots under the assumption of block attenuation of channel model, its purpose is to convert the signal $E[|s_k(t)|^2] = 1$. Next, the signal received by the satellite can then be expressed in the following

$$y_k(t) = \sum_{k=1}^K (\mathbf{H}_{US} + \mathbf{H}_{RS}\Phi\mathbf{h}_{UR})\mathbf{w}_k s_k + n_0 \quad (5)$$

105 where \mathbf{w}_k means the transmit power matrix coefficient vector at the k -th UEs of the
106 total transmit beamforming matrix \mathbf{W} , n_0 denotes the system noise followed by $n_0 \sim$
107 $\mathcal{CN}(0, \sigma^2)$, respectively.

108 2.1. Problem Formulation

It can be seen from Eq.(5) that the RIS-assisted IS-HAP-TNs system dose not introduce extra noise compared with the conventional relay-assisted system. This is because the RIS does not need to decode and encode the signal, but only acts as a simple reflective device and reflects the signal incident on it. The overall system sum rate is given by the following formula

$$R_k = \log_2(1 + \gamma_k) \quad (6)$$

where γ_k is the received signal-to-interference-plus-noise ratio (SINR), which can be shown as

$$\gamma_k = \frac{|(\mathbf{H}_{US} + \mathbf{H}_{RS}\Phi\mathbf{h}_{UR})\mathbf{w}_k|^2}{\sum_{j \neq k}^K |(\mathbf{H}_{US} + \mathbf{H}_{RS}\Phi\mathbf{h}_{UR})\mathbf{w}_j|^2 + \sigma^2} \quad (7)$$

The system can be described that the K UEs transmits signals to satellite, so its network performance should be the sum rate of the total K UEs, which can be modeled as

$$C(\mathbf{H}_{US}, \mathbf{H}_{RS}, \Phi, \mathbf{w}, \mathbf{h}_{UR}) = \sum_{k=1}^K R_k \quad (8)$$

109 Unlike traditional deep neural networks (DNNs), which require two phases, online
110 learning phase and offline training phase, each CSI is used to set up the states by our
111 proposed DRL approach, and the algorithm is used to obtain a continuous two matrices
112 through calculation. Mathematically speaking, the problem of RIS-assisted IS-HAP-TNs
113 performance optimization design can be represented as

$$\begin{aligned}
& \max_{\{\mathbf{W}, \Phi\}} C(\mathbf{H}_{US}, \mathbf{H}_{RS}, \Phi, \mathbf{w}, \mathbf{h}_{UR}) \\
& \text{s.t. } C_1 : \text{tr}\{\mathbf{W}\mathbf{W}^H\} \leq P_{\max} \\
& \quad C_2 : |\phi_m| \leq 1, \forall m = 1, 2, \dots, M.
\end{aligned} \tag{9}$$

114 where P_{\max} represents the maximum link transmission power. The constraint C_1 regu-
115 lates the UEs transmission maximum power. The constraint C_2 represents the constraints
116 on RIS reflective elements. Obviously, the above optimization problem are all based
117 on non-convex constraints and can hardly be settled by conventional improvement
118 approaches. If the classical mathematical tools are used, you must use exhaustive
119 exhaustive search to get a locally optimal or sub-optimal solution, which requires a
120 lot of computing resources or even impossible, especially for the large-scale network
121 communication scenarios proposed in this paper.

122 In each iteration of traditional alternating optimization algorithms, the globally
123 sub-optimal \mathbf{W} is solved by first fixing Φ and the sub-optimal Φ is solved by fixing the
124 matrix \mathbf{W} until the algorithm converges. For the design of high-dimensional continuous
125 variables, including the transmit power matrix, the phase shift matrix, etc., traditional
126 DRL methods such as DQN and DDPG cannot effectively solve these problems, and
127 often generate local optimal deviations.

128 3. Soft-DDPG-Based Joint Active and Passive Beamforming Design

129 In this section, the method of DRL is used to jointly optimize the transmit beam-
130 forming shape and phase shift array, and utilizing DDPG structure shown in Fig. 2.
131 First, we briefly discuss the soft-DDPG principle and operation process. Then, we will
132 introduce the proposed DNN architecture and provide a detailed description of the *state*,
133 *action*, *reward*, and the algorithm framework.

134 3.1. Overview of soft-DDPG

135 It is supposed that there exists a central controller or a learning agent in this network
136 that can collect channel information or communication date immediately, such as the
137 RIS to satellite channel \mathbf{H}_{RS} and \mathbf{h}_{UR} and the UE to the RIS channel \mathbf{h}_{UR} . Fig. 2 displays
138 the soft-DDPG architecture suitable for the earning agents to interact with high dynamic
139 communication environments to get pre-defined rewards or punishments. The core
140 concept of the soft-DDPG framework proposed in this letter is to perform effective
141 beamforming design and phase shift convert under unforeseen circumstances such as
142 local state observations such as RIS. The algorithm mainly includes two kinds of deep
143 neural networks (DNN), namely the training network and the target network. To avoid
144 or mitigate the issue of updating state participant values in a single case, we assume that
145 the target and training networks have the same neural network architecture.

146 Based on the above extensions, we can more clearly portray the framework covered
147 in this article, with four DNNs are drawn in detail, which are the training critic network,
148 the training actor network, the target critic network and the target actor network. The
149 functions of these four neural networks described above are described below. The
150 training critic network need to input the current state $s^{(t)}$ into the action network and
151 output the current action $a^{(t)}$, and the training actor network need to input the state $s^{(t)}$
152 and action $a^{(t)}$ into the training critic network and output the Q value $Q_{\pi}(s^{(t)}, a^{(t)})$. The
153 target critic network need to input the updated state $s^{(t+1)}$ to the target actor network
154 and output the $a^{(t+1)}$. The target actor network need to input the updated $s^{(t+1)}$ and
155 $a^{(t+1)}$ to the target critic network and output the target Q value $Q_{\pi}(s^{(t+1)}, a^{(t+1)})$.

156 Considering the existence of plural inputs in the neural network input, this pro-
157 posed model uses the *tanh* as the activation function of the hidden layer to limit the
158 action space in the interval $(0, 2\pi)$, and to eliminate the effect of the change in the distri-
159 bution of the hidden layer data brought by the parameter update. this proposed DRL
160 framework introduces a batch normalization layer after each hidden layer to process its

161 output. The batch normalization layer can effectively combat the gradient disappearance
 162 phenomenon, improve the training efficiency, and make the training process of the deep
 163 layer network more stable. In addition, according to the constraints of transmitting
 164 power and phase shift coefficients, the proposed model adds \tanh activation function
 165 to the output layer of the actor network to restrict the output to the interval $[-1,1]$, and
 166 subsequently transforms the action into the data format required by the optimization
 167 problem by taking absolute value normalization and range mapping methods to meet
 168 the constraints of power allocation and phase shift, so as to calculate the system sum
 169 rate as the Eq. (8).

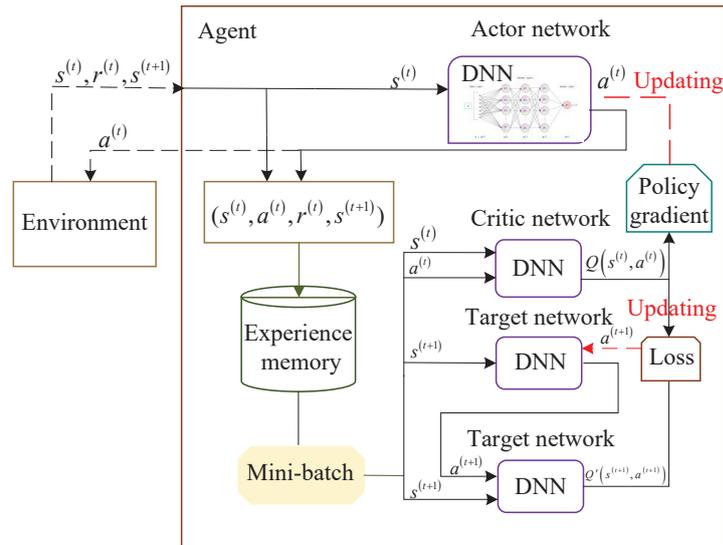


Figure 2. The DRL-based active transmit matrix and phase shift design framework using DDPG.

170 We generate different transmission link channel information by following the chan-
 171 nel model features described earlier when channel state information (CSI) and the
 172 previous action $\mathbf{W}^{(t-1)}$ and $\Phi^{(t-1)}$ are known at t -th time step, and the leaning agent
 173 can establish the knowledge about the current state space $s^{(t)}$ in the t -th time step. It
 174 is considered that the difficulty of joint optimal design of active transmission beamforming
 175 and passive RIS phase shift matrix are discrete and presents a great challenge to contin-
 176 uous state space and action space settings. Next, the detail of DRL-based algorithm state
 177 space S , action space A and the instant reward function R are explained below.

State: State space is generally a description of the environmental observations at t -th time step. In this paper, the DRL algorithm state space includes three parts, i.e., the last time action space, the satellite-RIS channel \mathbf{H}_{RS} and the RIS-UEs channel \mathbf{h}_{UR} . Next, the state of the t -th state space is defined as

$$s^{(t)} = [a^{(t-1)}, \mathbf{H}^{(t-1)}, \mathbf{h}_1^{(t-1)}, \dots, \mathbf{h}_K^{(t-1)}] \quad (10)$$

Action: Action space is generally a series of choices for the next action. Once the agent performs the current action $a^{(t)}$ step by step during the learning process according to the transfer policy π at the t -th time slot, the state space of the environment will be shifted from $s^{(t)}$ to the next state $s^{(t+1)}$. The actions $a^{(t)}$ are mainly related to the two variables \mathbf{W} and phase-shift matrix Φ to be optimized. Since Φ is a complex vector, to simplify the action dimension, this paper takes its phase part. Thus, the action space is modelled as

$$a^{(t)} = [\mathbf{w}_1^{(t)}, \dots, \mathbf{w}_K^{(t)}, \phi_1^{(t)}, \dots, \phi_M^{(t)}] \quad (11)$$

178 And, in the action space, to ensure that the neural network inputs are real numbers
 179 and match the neural network input formats, the variables to be optimized need to
 180 divide in real part and imaginary part, thus we define $\mathbf{W} = \text{Re}\{\mathbf{W}\} + \text{Im}\{\mathbf{W}\}$ and
 181 $\Phi = \text{Re}\{\Phi\} + \text{Im}\{\Phi\}$.

Reward: The purpose of this paper is to maximize the system sum rate and Eq. (10) is adopted as the reward function:

$$r^{(t)} = C(\mathbf{H}_{US}, \mathbf{H}_{RS}, \Phi, \mathbf{w}, \mathbf{h}_{UR}) \quad (12)$$

Algorithm 1: Soft-DDPG-based Algorithm

- 1: Initialize experience memory D to empty;
 - 2: Randomly initialization generate actor target/train network $\psi'(\cdot)$ and critic target/train network $Q'(\cdot)$ with parameters ζ'_a and ζ'_c , separately;
 - 3: **Input:** \mathbf{w} , ϕ , \mathbf{H}_{RS} and \mathbf{h}_{UR} ;
 - 4: **Output:** Optimal action $a_{opt}^{(t)}$;
 - 5: **for** each episode **do**:
 - 6: Initialize state $s^{(0)} \in S, S \leftarrow s^{(0)}$;
 - 7: **for** $t = 0, 1, 2, \dots, T - 1$ **do**:
 - 8: Choose action $a^{(t)} = \pi(s^{(t)} | \theta^\pi) + \mathcal{N}$;
 - 9: Take action $a^{(t)}$, get reward $r^{(t)}$ and $s^{(t)}$ evolves into new state $s^{(t+1)}$;
 - 10: Save $(s^{(t)}, a^{(t)}, r^{(t)}, s^{(t+1)})$ into D ;
 - 11: Randomly sample ζ transitions form D ;
 - 12: Traning framework via DNN;
 - 13: Compute target value for the critic's evaluation network by

$$y^{(i)} = r^{(i)} + \gamma Q'_{\pi'}(s^{(i+1)}, \pi'(s^{(i+1)} | \theta^{\pi'}) | \theta^{Q'})$$
 - 14: Update the parameters of the critic's evaluation network by

$$L(\theta^Q) = \frac{1}{\zeta} \sum_{i=1}^{\zeta} (y^{(i)} - Q_\pi(s^{(i)}, a^{(i)} | \theta^Q))^2;$$
 - 15: Update the parameters of actor network with sampled policy gradients by

$$\nabla_{\theta^\pi} J = \frac{1}{\zeta} \sum_{i=1}^{\zeta} \nabla_a Q_\pi(s, a | \theta^Q) \Big|_{a=\pi(s^{(i)} | \theta^\pi)} \nabla_{\theta^\pi} \pi(s | \theta^\pi);$$
 - 16: Soft-update the parameters of DDPG's target networks by

$$\theta_c^{(target)} \leftarrow \tau_c \theta_c^{(train)} + (1 - \tau_c) \theta_c^{(target)}$$

$$\theta_a^{(target)} \leftarrow \tau_a \theta_a^{(train)} + (1 - \tau_a) \theta_a^{(target)};$$
 - 17: Update the state $s^{(t+1)}$;
 - 16: **end for**;
 - 17: **end for**;
-

182

183 3.2. The Process of Algorithm Training

184 In order to break the coupling between experiences and adapt to a high dynamic
 185 environment, the experimencre replay approach allows agent access to previous historical
 186 experiences in subsequent training, the DDPG framework considered in this article.

187 For policy-class-based algorithms, the agent collects experience in episode. After run an
 188 episode, then lose your experience. Better with a multi-threaded parallel architecture.
 189 This not only solves the previous problems, but also makes efficient use of computing
 190 resources and improves training efficiency.

191 In the proposed DDPG framework, the entire agent consists of a global network and
 192 multiple parallel independent workers, each including a set of actor network and critic
 193 network. Each worker interacts independently with their own environment, gaining
 194 independent sampling experiences that are independent of each other, thus breaking the
 195 coupling between experiences to match the experience replay. Most of the underlying
 196 algorithms in DRL are single-threaded, that is, a learning agent that interacts with the
 197 environment to generate experience. Including the underlying version of actor network
 198 and critic network, because the environment is fixed and the action of the agent needs
 199 to be continuous, the experience gathered has strong timing associations and only part
 200 of the state and action space can be explored in a limited amount of time. To solve this
 201 problem, we adopt the DDPG scheme to optimize the design process and present the
 202 corresponding pseudo-code in Algorithm 1.

203 In the initial stage of the algorithm, the experience replay buffer D , the training
 204 actor network $\psi(\cdot)$ and training critic network $Q(\cdot)$ need to be initialized randomly (Line
 205 1-2). They are copied to the target network $\psi'(\cdot)$ and $Q'(\cdot)$ (Line 3). After initializing and
 206 randomly generating the RIS-assisted IS-HAP-TNs communication channel environment
 207 state, the state is processed via DNN and the output Y_t (Line 5-6). The action is derived
 208 based on Y_t , where \mathcal{N} is denoted as random noise, with the aim of seeking efficient
 209 exploration (Line 7-8). In this letter, we employ the mini batch to reduce the sample
 210 training amount of sampling and ensure the quality of gradient reduction. After the
 211 transformation sequence is saved in the memory replay buffer D (line 10), to achieve
 212 the optimal action that maximizes the output of the critic train network, the two train
 213 networks are updated using the minibatches of size ζ randomly sampled from replay
 214 buffer D (line 11). Update the critic target network parameters $Q(\cdot)$ by minimizing the
 215 variance loss (line 14). Make use of linking rules to update the actor networks parameters
 216 $\psi(\cdot)$. Finally, the target networks parameters of actor network and critic network are
 217 slowly soft updated using the control factor τ as the decaying rate (line 15).

During each iteration t each of the learning process, the actor train network will
 select the action from the continuous action space based on the current state $s^{(t)}$. During
 this training process, in order to effectively explore the optimal action, the stochastic
 noise \mathcal{N}_a is also taken into account in the algorithm framework to obtain the deterministic
 strategy, i.e., $a^{(t)} = \pi(s^{(t)} | \theta^\pi) + \mathcal{N}_a$, where θ^π is the actor train network parameter,
 and π is the transfer policy. When the operation ends, the environment will transit the
 last action to the next state $s^{(t+1)}$ to obtain instant reward $r^{(t)}$, and then get an evaluation
 for the action to evaluate the optimal action $a^{(t)}$. Modeling a state-action value function
 by parameterized by θ^Q as

$$Q_\pi(s^{(t)}, a^{(t)} | \theta^Q) \leftarrow \alpha Q_\pi(s^{(t)}, a^{(t)} | \theta^Q) + (1 - \alpha) \left[r^{(t)} + \gamma \max_{a'} Q_\pi(s^{(t+1)}, a' | \theta^Q) \Big|_{a' = \pi(s^{(t+1)} | \theta^\pi)} \right] \quad (13)$$

218 where α denotes the algorithm learning rate in this algorithm framework. To ensure
 219 the stability, the target actor network parameterized by $\theta^{\pi'}$ and the target critic network
 220 characterized by $\theta^{Q'}$, which is parameterized at intervals according to online network
 221 parameters. Thus, considering the parameter update strategy is a soft update method,
 222 so the algorithm is called soft-DDPG. The soft update method of parameters ensures the
 223 slow update of parameters and alleviates the instability problem of the policy network
 224 during the learning process.

225 4. Numerical Simulation Results

226 In this section, we will evaluate the performance improvement of the proposed
 227 DRL-based algorithm framework for the proposed system model from different perspec-
 228 tives.. First, we will randomly generate channel model matrix \mathbf{H}_{RS} and \mathbf{h}_{UR} following
 229 shadowed-Rician fading distribution and Rayleigh distribution, respectively [25]. The
 230 system parameters and hyperparameters of the DDPG algorithms are listed in Table
 231 1 [26]. To test whether our algorithm improves network performance, we will also
 232 consider three other standard solutions:

233 1) Hard-DDPG: The scheme indicates that the parameters in the DRL framework
 234 are updated in a hard-update strategy which that allows the network to copy all the
 235 parameters in the network at this time directly into the target network after every t_u
 236 training sessions by pre-setting the parameter update interval t_u .

237 2) Random RIS: The scheme denotes that the RIS phase shift matrix Φ is randomly
 238 generate.

239 3) Without RIS: This scheme denotes that the communication scenario without RIS
 240 and the UEs can send the signals directly to the satellite. Considering that the process is
 241 a continuous transmission, thus we assume that the successful transmission signal is
 242 $1/2$.

243 Fig. 3 plots the relationship between the number of RIS reflection elements and the
 244 system sum rate. As can be seen from the figure, the system sum rate is significantly
 245 higher for all algorithms as the number of RIS elements increases, due to the fact that
 246 more RIS reflection elements increase the reflection channel gain, but also sacrifices the
 247 complexity of the RIS deployment at the HAP. In addition, we can observe that the
 248 soft-update parameter strategy obtains a higher system sum rate than the hard update
 249 parameter strategy, alleviates the instability of the Q-value network in the learning
 250 process, and the soft-update strategy obtains a higher system ensemble rate by more
 251 flexible interacting with the environment to design the phase shift matrix more flexibly.

252 The setting of hyper-parameters will have a great impact on the performance, like
 253 the stability and convergence speed of neural networks. This paper also explores the
 254 effect of different learning rates on the performance and convergence speed of the
 255 model in our proposed DRL framework. The average reward is used to measure its
 256 performance, which can be shown as

$$\text{average_reward}(T_i) = \frac{\sum_{t=1}^T r^{(t)}}{T_i}, T_i = 1, 2, \dots, T \quad (14)$$

257 where T is the maximum step size of sample training. Fig. 4 shows the average reward
 258 versus time step under different learning rates, and it can be seen that the effect of
 259 different learning rate settings in the neural network on the performance of the DRL
 260 algorithm varies greatly. In particularly, the considered DRL framework with a learning
 261 rate of 0.001 performs the best, but converges more slowly than the others. As the
 262 RIS reflection element increases, the average system reward also increases gradually as
 263 expected with the addition of reflection channels, but this does not significantly increase
 264 the convergence time of the proposed DRL framework.

265 Fig. 5 shows a schematic comparison of the average reward performance and the
 266 outdated CSI coefficients, respectively. In the proposed DRL framework, we choose the
 267 last moment CSI as the state space input, and we can see that the average reward of all
 268 algorithms decreases gradually as the outdated CSI coefficient decreases. However, the
 269 proposed soft-DDPG framework remains at a favorable level compared to the existing
 270 scheme and the hard-DDPG scheme. Compared with the advanced DRL schemes, which
 271 do not require an exact channel model information, the existing alternating parameter
 272 optimization scheme relies on the knowledge of static exact channel model, but because
 273 of the high dynamic communication scenario, the system performance is not as good as
 274 soft DDPG and hard DDPG scheme.

Table 1: System and DNN Parameters

System parameters	Value
Frequency band	$f = 2$ GHz
Wavelength	$\lambda = 150$ mm
Noise power spectral density	-169 dBm/Hz
Link bandwidth	$W = 15$ MHz
Noise temperature	$T = 300$ K
Height of HAP	20km
Number of the UEs	$K=3$
Transmission path	$L=3$
DNN hyperparameters in DDPG	
Reward discount rate	0.99
Numbers of experiences with the mini-batch	16
Learning rate	0.0001
Decaying rate	0.0001
Experience replay buffer size	100000
Numbers of steps in each training episode	10000

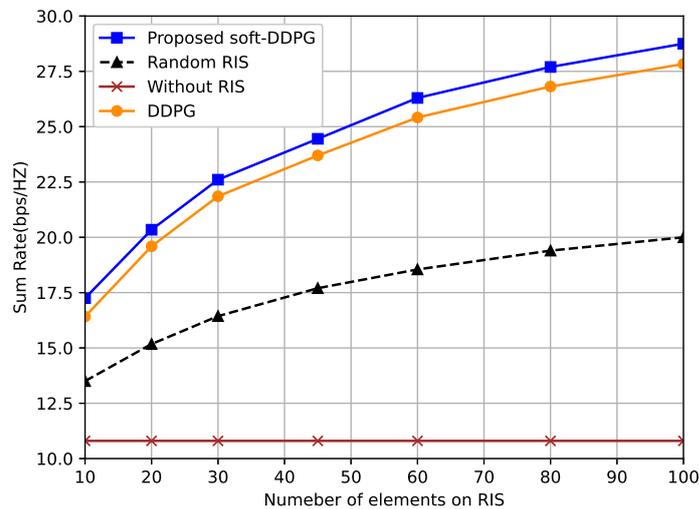


Figure 3. Sum rate performance relative to the increasing number of elements on RIS.

275 5. Conclusion

276 This paper discussed the joint optimal design scheme of transmitting active beam-
 277 forming and passive beamforming for maximizing system sum rate. In the IS-HAP-TNs
 278 assisted by RIS, it is hard to sense the channel state information in the dynamic environ-
 279 ment accurately and comprehensively. On this basis, a novel type of DRL architecture,
 280 namely soft-DDPG algorithm. With the help of the network parameter soft-update strat-
 281 egy, the coordination of the phase shift matrix can be obtained even when the increasing
 282 number of RIS reflective elements amplitude changes. Simulation results show that
 283 the proposed framework can achieve better network performance in a lower operation
 284 duration and can be applied to the real-time control of IS-HAP-TNs system.

285 **Acknowledgments:** This work was supported by the National Science Natural Foundation of
 286 China (No.61901502, No.61971474 and No.62001517), the National Postdoctoral Program for
 287 Innovative Talents (No.BX20200101), the Research Project of Science and Technology on Com-
 288 plex Electronic System Simulation Laboratory (No.DXZT-JC-ZZ-2019-005), and in part by Bei-
 289 jing Nova Program (No.Z201100006820121) and the Project of Space Engineering University
 290 (No.2020XXAQ01 and No.2019XXAQ05)

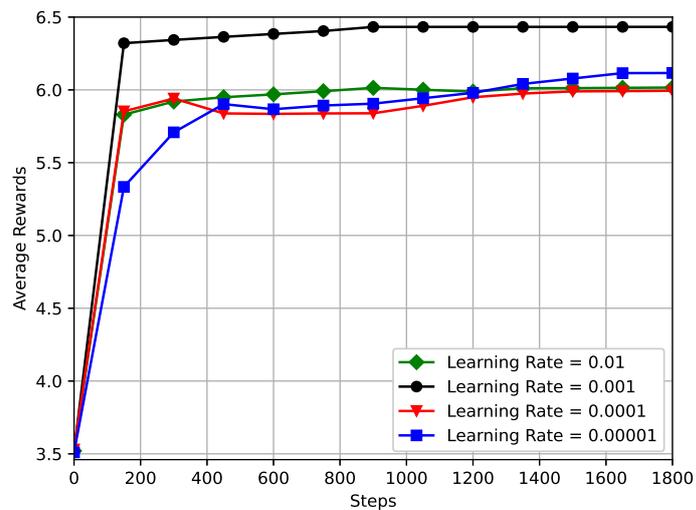


Figure 4. Variation of average reward under different learning rate.

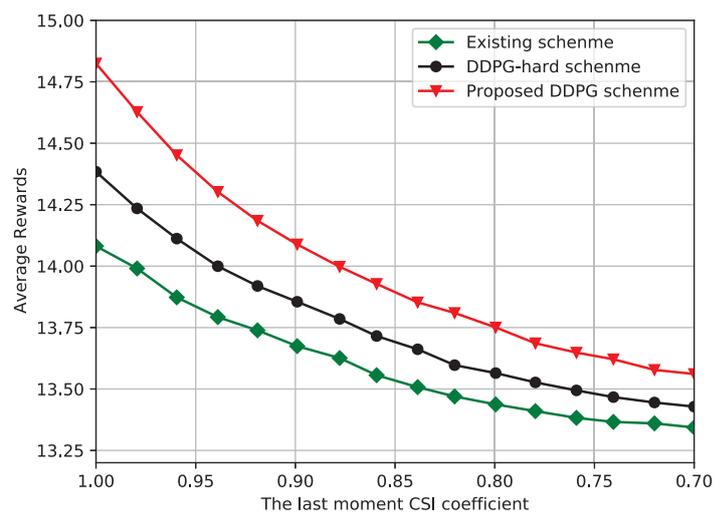


Figure 5. Average reward against the last moment CSI coefficient.

291 **Author Contributions:** Min Wu, Shibing Zhu, Changqing Li, Yudi. Chen, Feng Zhou and Xiang
 292 Su conceived and designed the experiments; Min Wu performed the experiments; Shibing Zhu,
 293 Changqing Li and Yudi. Chen analyzed the data; Feng Zhou and Xiang Su contributed analysis
 294 tools; Min Wu, Shibing Zhu and Feng Zhou wrote the paper.

295 **Conflicts of Interest:** The authors declare no conflict of interest.

296 References

- 297 1. K. An, M. Lin, J. Ouyang, and W. P. Zhu, "Secure transmission in cognitive satellite terrestrial
 298 networks," *IEEE J. Sel. Areas Commun.*, vol. 34, no. 11, pp. 3025-3037, Nov. 2016.
- 299 2. Lin, Z., Lin, M., Cola, T. de, Wang, J.-B., Zhu, W.-P., Cheng, J. "Supporting IoT with rate-
 300 splitting multiple access in satellite and aerial integrated networks," *IEEE Internet Things J.*,
 301 **2021**, *8*, 11123-11134.
- 302 3. Liu, R., Guo, K., An, K., Zhu, S., Shuai, H. "NOMA-based integrated satellite-terrestrial
 303 relay networks under spectrum sharing environment," *IEEE Wireless Commun. Lett.*, **2021**, *10*,
 304 1266-1270.

- 305 4. L. Yang, P. Li, Y. Yang, S. Li, I. Trigui and R. Ma, "Performance Analysis of RIS-Aided
306 Networks With Co-Channel Interference," *IEEE Commun. Lett.*, vol. 26, no. 1, pp. 49-53, Jan.
- 307 5. H. Luo, L. Lv, Q. Wu, Z. Ding, N. Al-Dhahir and J. Chen, "Beamforming Design for Active
308 IOS Aided NOMA Networks," *IEEE Wireless Commun. Lett.*, 2022.
- 309 6. H. Niu, Z. Chu, F. Zhou, et al., "Robust design for intelligent reflecting surface-assisted
310 secrecy SWIPT network," *IEEE Trans. Wireless Commun.*, vol. 21, no. 6, pp. 4133-4149, June.
311 2022.
- 312 7. S. Gong, X. Lu, D. T. Hoang, et al., "Towards smart wireless communications via intelligent
313 reflecting surfaces: A contemporary survey," *IEEE Commun. Surv. Tut.*, pp. 1-33, Jun. 2020.
- 314 8. X. Li et al., "Hardware impaired ambient backscatter NOMA systems: Reliability and secu-
315 rity," *IEEE Trans. Commun.*, vol. 69, no. 4, pp. 2723-2736, April 2021.
- 316 9. K. Guo, K. An, et al., "Physical layer security for multiuser satellite communication systems
317 with threshold-based scheduling scheme," *IEEE Trans. Veh. Technol.*, vol. 69, no. 5, pp. 5129-
318 5141, May. 2020.
- 319 10. J. Wang, Y.-C. Liang, et al., "Robust beamforming and phase shift design for IRS-enhanced
320 multi-user MISO downlink communication," *Proc. IEEE Int. Conf. Commun. (ICC)*, Dublin,
321 Ireland, pp. 1-6, Jun. 2020.
- 322 11. K. B. Letaief, W. Chen, et al., "The roadmap to 6G: AI empowered wireless networks," *IEEE*
323 *Commun. Mag.*, vol. 57, no. 8, pp. 84-90, Aug. 2019.
- 324 12. X. Yue and Y. Liu, "Performance Analysis of Intelligent Reflecting Surface Assisted NOMA
325 Networks," *IEEE Trans. Wireless Commun.*, vol. 21, no. 4, pp. 2623-2636, April 2022.
- 326 13. Q.-U.-U. Nadeem, A. Kammoun, A. Chaaban, M. Debbah, and M.-S. Alouini, "Asymptotic
327 max-min SINR analysis of reconfigurable intelligent surface assisted MISO systems," 2019,
328 *arXiv:1903.08127*. [Online]. Available: <http://arxiv.org/abs/1903.08127>.
- 329 14. M. Fozzi, A. R. Sharafat and M. Bennis, "Fast MIMO beamforming via deep reinforcement
330 learning for high mobility mmWave connectivity," *IEEE J. Sel. Areas Commun.*, vol. 40, no. 1,
331 pp. 127-142, Jan. 2022.
- 332 15. Y. Li, W. Zhang, C.-X. Wang, et al., "Deep reinforcement learning for dynamic spectrum
333 sensing and aggregation in multi-channel wireless networks," *IEEE Trans. Cognitive Commun.*
334 *Net.*, vol. 6, no. 2, pp. 464-475, June. 2020.
- 335 16. H. Ren, C. Pan, L. Wang, et al., "Long-term CSI-based design for RIS-aided multiuser MISO
336 systems exploiting deep reinforcement learning," *IEEE Commun. Lett.*, vol. 26, no. 3, pp.
337 567-571, March. 2022.
- 338 17. M. Zhang, S. Fu and Q. Fan, "Joint 3D deployment and power allocation for UAV-BS: A deep
339 reinforcement learning approach," *IEEE Wireless Commun. Lett.*, vol. 10, no. 10, pp. 2309-2312,
340 Oct. 2021.
- 341 18. F. B. Mismar, B. L. Evans, and A. Alkhateeb, "Deep reinforcement learning for 5G networks:
342 Joint beamforming, power control, and interference coordination," *IEEE Trans. Commun.*, vol.
343 68, no. 3, pp. 1581-1592, Mar. 2020.
- 344 19. R. Shafin, M. Jiang, S. Ma, L. Piazzzi and L. Liu, "Joint Parametric Channel Estimation and Per-
345 formance Characterization for 3D Massive MIMO OFDM Systems," *2018 IEEE International*
346 *Conference Commun. (ICC)*, 2018, pp. 1-6.
- 347 20. C. Huang, R. Mo and C. Yuen, "Reconfigurable Intelligent Surface Assisted Multiuser MISO
348 Systems Exploiting Deep Reinforcement Learning," *IEEE J. Sel. Areas Commun.*, vol. 38, no. 8,
349 pp. 1839-1850, Aug. 2020.
- 350 21. Liang Yang, Qi Zhu, Sai Li, Imran Shafique Ansari, and Siyuan Yu, "On the Performance of
351 Mixed FSO-UWOC Dual-Hop Transmission Systems," *IEEE Wireless Commun. Lett.*, vol.10,
352 no.9, pp.2041-2045, Sep.2021.
- 353 22. Yang, F. Meng, Q. Wu, D. B. da Costa and M. -S. Alouini, "Accurate Closed-Form Approxi-
354 mations to Channel Distributions of RIS-Aided Wireless Systems," *IEEE Wireless Commun.*
355 *Lett.*, vol. 9, no. 11, pp. 1985-1989, Nov. 2020.
- 356 23. J. He, M. Leinonen, H. Wymeersch, et al., "Channel estimation for RIS-aided mmWave MIMO
357 systems," in *Proc. 2020 IEEE Global Commun. Conf. (GLOBECOM)*, Taipei, Taiwan, pp. 1-6,
358 Dec. 2020.
- 359 24. Yan, X.; Xiao, H.; An, K.; Zhen, G.; Chatzainotas, S. "Ergodic capacity of NOMA-based uplink
360 satellite networks with randomly deployed users," *IEEE Syst. J.*, 2020, 14, 3343-3350.
- 361 25. Bletsas, A.; Shin, H.; Win, M.Z. "Cooperative communication with outage-optimal oppor-
362 tunistic relaying," *IEEE Trans. Wireless Commun.*, 2007, 6, 3450-3460.

- 363 26. K. Guo, M. Lin, J.-B. Wang, et al., "On he performance of LMS communication with hardware
364 impairments and interference," *IEEE Trans. Commun.*, vol. 67, no. 2, pp. 1490-1505, Feb. 2019.

