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Hybrid Carrier Frequency Modulation Based on Rotor Position to Reduce Sideband Vibro-acoustics in PMSM Used by Electric Vehicle

Zizhen Qiu¹, Yong Chen^{1*}, Xiaozhe Lin², Haiquan Cheng¹, Yang Kang¹, Xu Liu³

¹ Tianjin Key Laboratory of Power Transmission and Safety Technology for New Energy Vehicles, Hebei University of Technology, Tianjin 300130, China.

qiuzizhen@163.com; CHQ9633@163.com; kangyang1125@163.com.

² Zhejiang Geely Automobile Co., Ltd.

joelin@geely.com.

³ State Key Laboratory of Reliability and Intelligence of Electrical Equipment, Hebei University of Technology, Tianjin 300130, China.

liuxu@hebut.edu.cn.

* Correspondence: chenyong1585811@163.com; Tel.: +86-15858118288.

Abstract: In the permanent magnet synchronous motor (PMSM) drive system, the unwilling and ear-piercing vibro-acoustics caused by high-frequency sideband harmonics becomes unacceptable in the electric vehicle application. In this paper, a modified space vector pulse-width modulation (SVPWM) technique implemented with hybrid carrier frequency modulation (HCFM) is provided to reduce the sideband current harmonic components and vibro-acoustic responses. The principle and implementation of the proposed HCFM technique are firstly presented, in which the fixed carrier frequency is improved with the sawtooth and random signal-based coupling modulation based on the rotor position. For verification, the experiment tests are carried out on a prototype 12/10 PMSM and microcontroller unit. The effectiveness of the HCFM technique can hence be confirmed, in which the sideband vibro-acoustics reduction shows more effectively than that in conventional random PWM. The proposed approach may provide a new route in noise-cancelling and electromagnetic compatibility for the electric drive powertrain.

Keywords: permanent magnet synchronous motor; sideband harmonic component; space vector pulse-width modulation; carrier frequency modulation; vibro-acoustic responses.

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1. Introduction

In recent years, permanent magnet synchronous motor (PMSM) driven with pulse-width modulation (PWM) techniques have widely used in electric vehicles (EVs) and hybrid electric vehicles (HEVs) **Error! Reference source not found..** The high-frequency vibro-acoustics responses introduced by sideband voltage/current harmonics are mainly located and contributed around the carrier frequency and its multiples, which brings much more noise, vibration and harshness (NVH) challenges during the design and optimization stages of the vehicle platform **Error! Reference source not found..**

Which has been confirmed that electromagnetic vibro-acoustics is the most significant contributor to the PMSM system due to electromagnetic forces on the stator and rotor caused by air-gap magnetic fields, such as the cogging torque **Error! Reference source not found..**, torque ripple **Error! Reference source not found..** and the high-frequency sideband vibro-acoustics **Error! Reference source not found..** In contrast to the previous two types, the sideband vibro-acoustics is related to the control unit, associated with the

voltage source inverter (VSI), and also to the pulse-width modulation (PWM) technique. The fixed carrier frequency causes the current harmonic components to contribute around the carrier frequency and its multiples **Error! Reference source not found.** With the interaction between the permanent magnet (PM) field and the armature reaction field, the ear-piercing vibro-acoustic responses can thus be produced **Error! Reference source not found.**

In order to reduce the sideband components, some effective methods have been presented [10-14]. The most representative method is called the harmonics spread spectra (HSS) technique, such as periodic signal-based dithering modulation **Error! Reference source not found.** and random pulse-width modulation (RPWM) **Error! Reference source not found.** The periodic dithering modulation transfers the fixed carrier frequency to the determined periodic ways, such as sawtooth, sinusoidal, and square **Error! Reference source not found.** Ditto for RPWM, the carrier frequency would be randomly distributed within a defined range **Error! Reference source not found.** However, the related single-HSS technique would exhibit saturating and limiting effect such that the sideband components cannot be further reduced **Error! Reference source not found.**, even other negative influences, increasing the total harmonic distortion (THD) **Error! Reference source not found.** and electromagnetic interference (EMI) **Error! Reference source not found.**, on the power converter system.

Therefore, this paper presents a novel hybrid carrier frequency modulation (HCFM) method to modify the conventional space vector pulse-width modulation (SVPWM) technique, in which the fixed carrier frequency can be dithered by coupling the sawtooth and random signal-based patterns with different rotor position angles. The corresponding experimental tests are presented with great practical significance to verify the proposed HCFM technique.

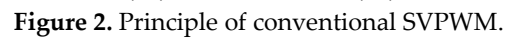
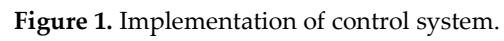
The paper is organized into five sections. Following this introduction, the principle and implementation of the HCFM technique are presented in Section II, where the explanation of sideband harmonics reduction is also proposed. The experimental setup is presented in Section III. Then, the results are obtained in Section IV. Finally, the conclusion is summarized in Section V.

2. Hybrid Carrier Frequency Modulation

2.1 Principle and Implementation of the HCFM

The implementation of the conventional SVPWM is shown in Figure 1 and Figure 2. Taking Sector I for an example, the output PWM signals, q-axes voltages and q-axes current are shown in Figure 3. With the d-q voltage calculated models, harmonic analysis can be developed with quantitative analysis of the q-axes voltage, which can be confirmed that the sideband harmonic components are related to the time duration of the zero-vectors T_0 and T_7 . Moreover, the time durations are also related to the rotor position **Error! Reference source not found.** Hence, the hybrid carrier frequency modulation is based on the rotor position to combine the corresponding periodic and random signal-based dithering methods, which can be seen in Figure 4.

As shown in Figure 4, the carrier frequency in this work is set at 8000Hz. The periodic signal, associated with a sawtooth form, is employed, as while the regular random signal is also employed. The dithering range is set as 1000Hz, which means the modified carrier frequency is distributed between 7000Hz and 9000Hz. Changing with the rotating angles of the rotor, the carrier frequency dithers per 30 degrees, in which the triangle waves of the carrier frequency are also simply presented.



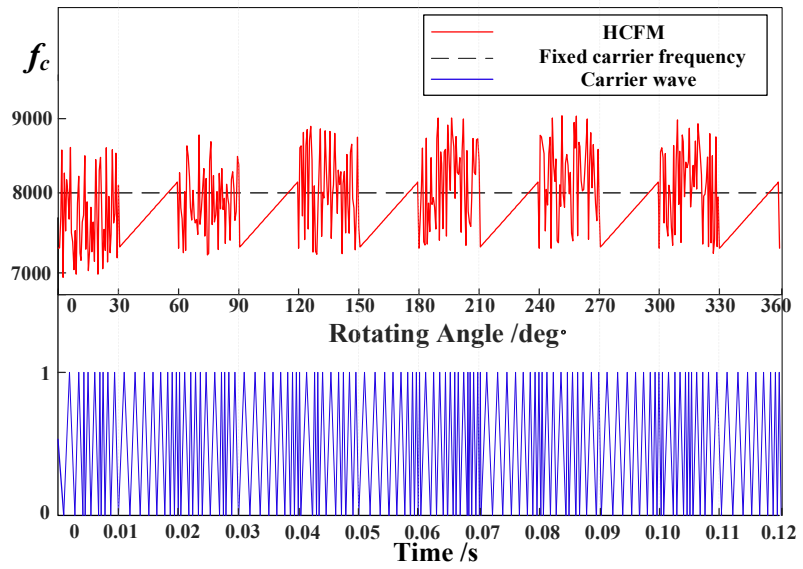


Figure 4. The variation laws and the corresponding carrier waves of carrier frequency with HCFM.

2.2 Explanation of Sideband Harmonics Reduction

The analysis of the sideband harmonics can be proposed by the quantitative analysis of q-axis voltages. Neglecting the stator resistance, the working angle in the steady-state can be expressed as δ . Acting with U_4 and U_6 separately, U_q can be expressed respectively as **Error! Reference source not found.**

$$U_{4q} = \frac{2}{3} U_{dc} \cos(\theta - \delta) \quad (1)$$

$$U_{6q} = \frac{2}{3} U_{dc} \cos(\delta - \theta + 60^\circ) \quad (2)$$

$$\vec{U}_q = \frac{T_4}{T_s} \vec{U}_{4q} + \frac{T_6}{T_s} \vec{U}_{6q} \quad (3)$$

where, T_4 and T_6 can be expressed by using the phase voltage amplitude U_m and the vector clamping angle θ .

$$T_4 = \sqrt{3} \frac{U_m}{U_{dc}} T_s \sin(60^\circ - \theta) \quad (4)$$

$$T_6 = \sqrt{3} \frac{U_m}{U_{dc}} T_s \sin \theta \quad (5)$$

$$\begin{aligned} T_0 &= T_s - T_4 - T_6 \\ &= T_s \left[1 - \frac{\sqrt{3} U_m}{U_{dc}} \cos(\theta - 30^\circ) \right] \end{aligned} \quad (6)$$

With (1), (2) and (6), the phase wave in one period can be obtained in Figure 3. Then, the first and second orders of current harmonics, Δi_{q1} and Δi_{q2} , can be presented as

$$\Delta i_{q1} \approx \frac{U_m T_s \cdot \left[\sin(60^\circ + \delta) - \sin \frac{\delta}{2} \right]}{4\sqrt{3}L_q} \quad (7)$$

$$\Delta i_{q2} \approx \frac{U_q T_s}{4L_q} \cdot \left[1 - \frac{\sqrt{3}U_m}{2U_{dc}} \right] \quad (8)$$

It can be seen from (7) and (8) that the second order of current harmonics mainly affects q-axis current harmonic components. Moreover, the magnitude of Δi_q is determined by T_0 , L_q and U_q . In this study, the T_0 can thus be optimized by the rotor position to reduce the sideband harmonics. The optimized configuration can be shown in Figure 4 and expressed as

$$f_s = \begin{cases} f_{t1} + R_i \Delta f & \left[\frac{k\pi}{6} < \theta < \frac{(k+1)\pi}{6} \right] \\ f_c + f_{t2} \Delta f & \left[\frac{(k+1)\pi}{6} < \theta < \frac{(k+2)\pi}{6} \right] \end{cases} \quad (9)$$

In (9), f_c is the carrier frequency. Δf is the dithering index defining the harmonics spread range, which sets 1000Hz in this study. f_{t1} is the frequency of the sinusoidal wave. f_{t2} is the frequency of sawtooth wave meeting as $[-1, 1]$. R_i is the random modulation index. According to the above analysis, the novel HCFM based on rotor position not only reduces harmonic currents but also achieves a wider spectrum extension.

3. Experimental Test Setup

As shown in Figure 5 and Figure 6, the test bench, set up by prototype PMSM, power-driven system, and measurements, is presented to verify the effectiveness of the proposed HCFM technique. The prototype 12/10 PMSM is adopted to the rear axle driven unit in an EV application, whose parameters can be found in Table 1. The DC power is fed by 540V-4.5kw. Three full-bridge power modules are employed by circuit boards with Infineon-BSM75GB120DN2. Based on the dSPACE-MicroLabBox 1103, serval PWM strategies can be established with MATLAB/Simulink and RTI Electric Motor Control Blockset. The corresponding pulse signals can be generated and controlled online.

The vibro-acoustic tests are measured by ICP triaxial accelerometer with a sensitivity of 50 mV/g and a sound-pressure microphone (GRAS-46AE) with the sensitivity of 47.23mV/Pa. The phase current signals are recorded by Teletronix-A622 current probe and processed by Tektronix TDS2024c.

The operational condition of all experimental tests is set at 1000 r/min and 4 Nm, in order to obtain the sideband current harmonics and vibro-acoustic responses obviously. Based on the setting rotating speed 1000 r/min, the rotation frequency and fundamental frequency are 16.67 Hz and 83.34 Hz, respectively.

Table 1. Parameters of prototype PMSM

Item	Value	Item	Value
Number of slots	12	Type of controller	VSI
Number of poles	10	Carrier frequency	8000 Hz
Rated speed	2000 r/min	DC link voltage	540 V
Rated torque	8 Nm	Rated power	1.8 kW

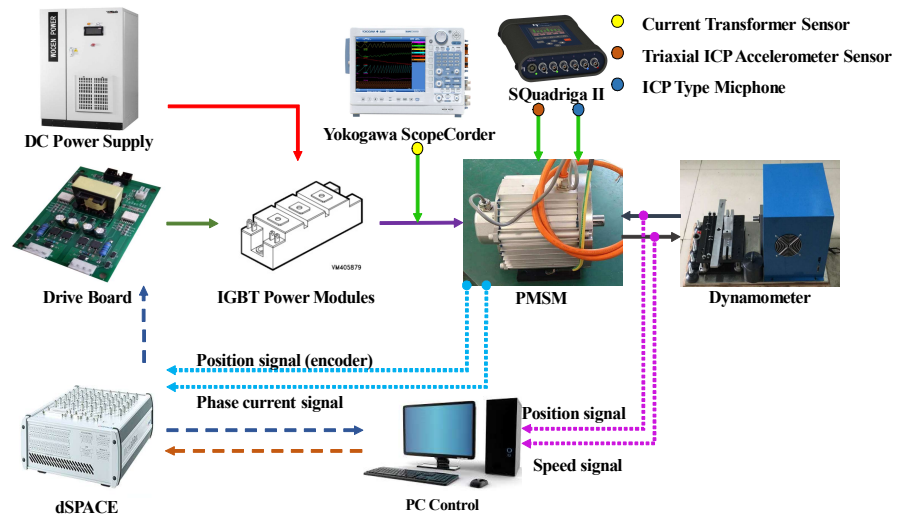


Figure 5. Overview of the test platform.

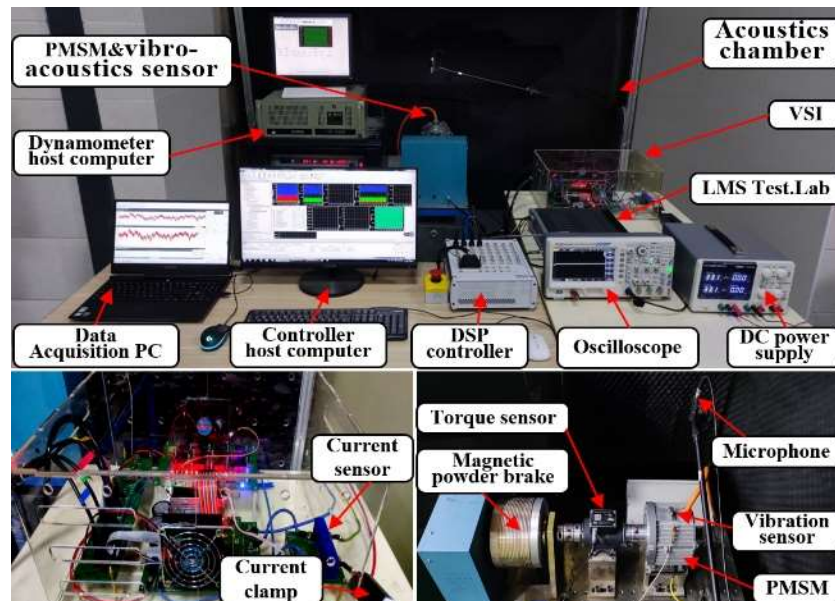


Figure 6. Experimental setup with the PMSM and instrumentation

4. Results and Discussions

4.1 Sideband Current Harmonics

The experimental results of sideband current harmonics are shown in Figure 7, in which the comparison among the conventional SVPWM, RPWM and HCFM techniques is presented by the power spectrum density (PSD). The sideband components in conventional SVPWM obtain a specific pattern in that two symmetric sidebands appear around the first carrier frequency. The sidebands at $f_c \pm 2f_0$ and $f_c \pm 4f_0$, where f_0 is the fundamental frequency of the electrical supply, which can be calculated by the rotation frequency f_r , i.e., $f_r = n/60$ and $f_0 = p \cdot f_r$. Considering that f_0 is 83.34Hz under 1000r/min, the main sidebands can be obtained with four main orders, where the peak magnitude is -27.56 dB/Hz.

With the RPWM and HCFM, the current harmonic reduction can be seen in Figures 8 and Figure 9. Different from the conventional SVPWM, the obvious orders are disappeared. The concentrated harmonics are extended to the specified frequency domain with a 1000Hz dithering range as the setting. The peak magnitude in RPWM is -39.56 dB/Hz,

while -48.78 dB/Hz or even below in HCFM. The reduction effort in the proposed HCFM is more significant than that in RPWM.

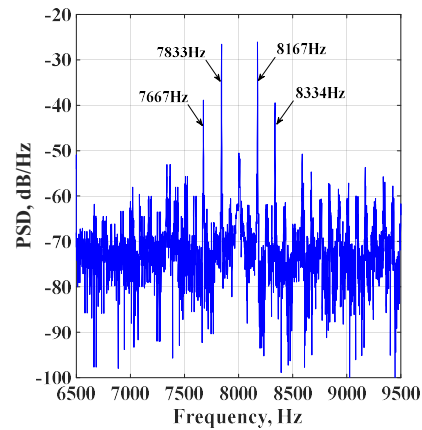


Figure 7. The sideband current harmonics with conventional SVPWM

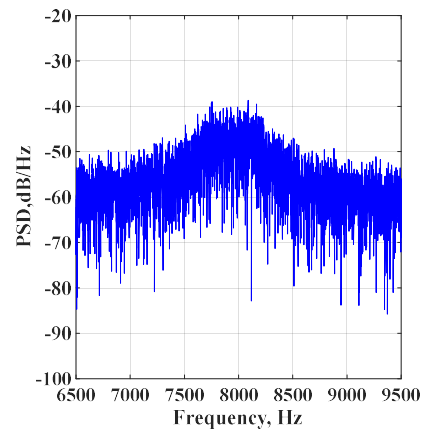


Figure 8. The sideband current harmonics with random PWM

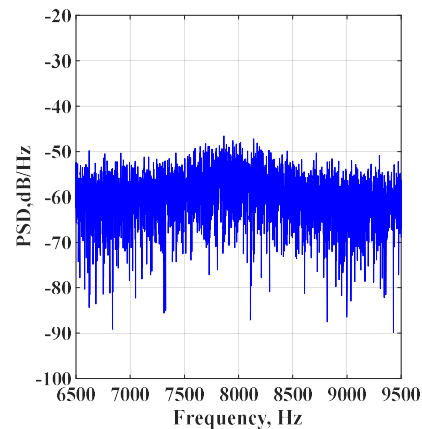


Figure 9. The sideband current harmonics with HCFM

4.2 Sideband Vibro-acoustic Responses

The experimental vibro-acoustics results are carried out to the frequency domain. The comparison of the sideband vibro-acoustic responses is shown in this chapter. As shown in the following Figures, the vibro-acoustic responses obtain a significant correlation with the current harmonics. In Figure 10, the sideband orders of the conventional SVPWM in vibro-acoustic responses are located with $f_{\pm f_0}$, $f_{\pm 3f_0}$, and $f_{\pm 5f_0}$.

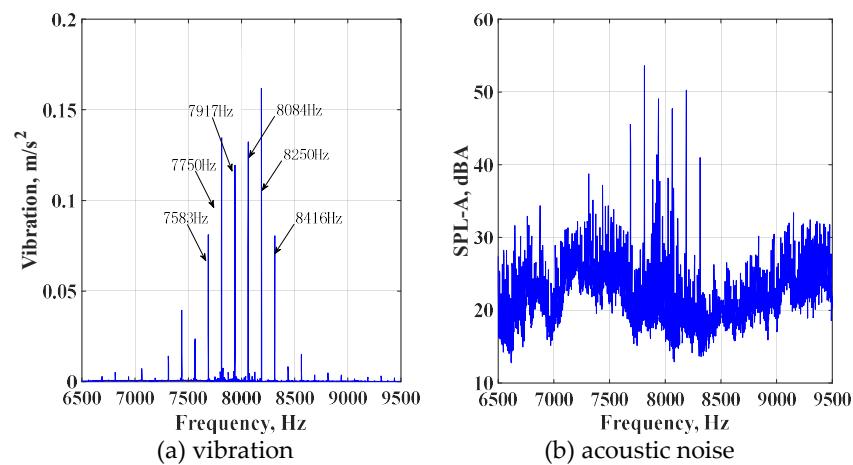


Figure 10. The sideband vibro-acoustic responses with conventional SVPWM

With the reduction techniques in Figure 11 and Figure 12, the relatively concentrated sideband components are spread to a wider frequency band, which has been cut off at lower and upper frequencies 7000 Hz and 9000 Hz. The original narrow-band acoustics are reduced below 50 dBA, even below 40 dBA by using HCFM.

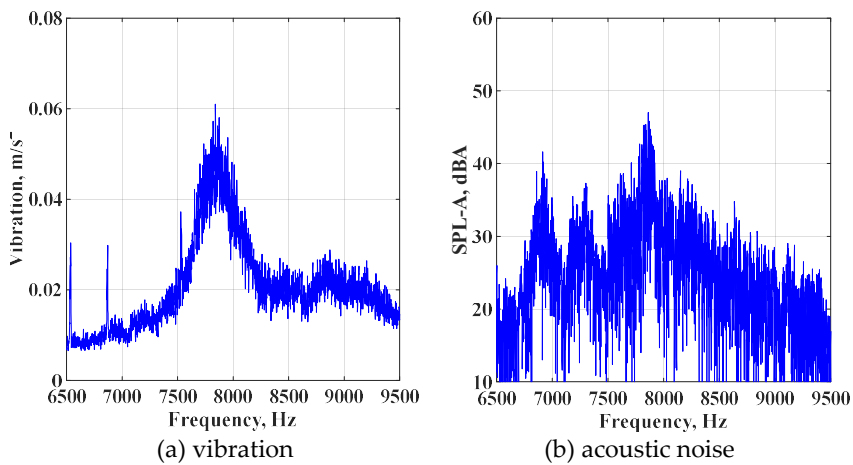


Figure 11. The sideband vibro-acoustic responses with random PWM

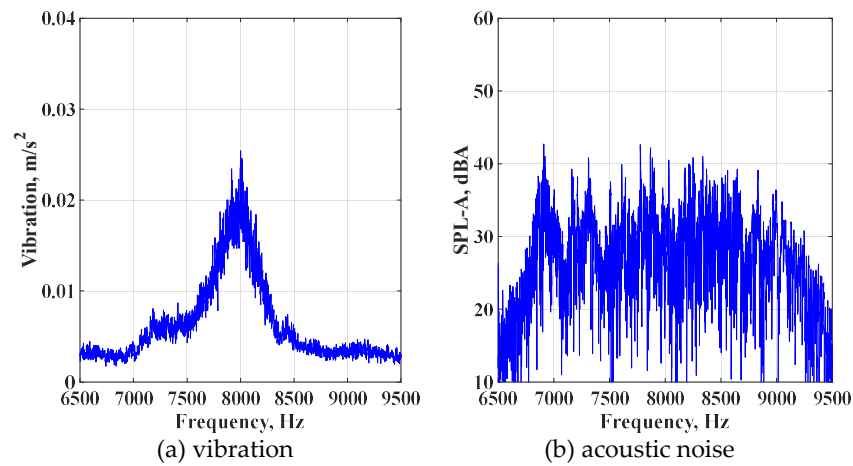


Figure 12. The sideband vibro-acoustic responses with HCFM

5. Conclusions

In this paper, a novel PWM technique applied with hybrid carrier frequency modulation is proposed to reduce the sideband harmonics and vibro-acoustics. The proposed HCFM technique is implemented by the rotor position, in which sawtooth and random signal-based dithering methods are combined to achieve the hybrid carrier waves modulation. With the experimental tests, the magnitude of sideband current harmonics can be reduced by about 20 dB/Hz, resulting in a 0.12 m/s² vibration reduction and 10 dBA acoustic noise reduction. The reduction efforts in HCFM are more significant than those in RPWM. The results can provide a reference for multi-strategies control to achieve further vibro-acoustics and EMI reduction in EV application.

Author Contributions: All authors contributed to this work. Conceptualization, Zizhen Qiu., Formal analysis, Haiquan Cheng, Investigation, Zizhen Qiu, Yong Chen and Xu Liu, Methodology, Zizhen Qiu and Haiquan Cheng, Validation, Zizhen Qiu and Yang Kang, Writing—original draft, Zizhen Qiu, Writing—review and editing, Yong Chen and Xu Liu., Funding acquisition, Xiaozhe Lin.

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Conflicts of Interest: The authors declare no conflict of interest.

References

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