

1 *Review*

## 2 **Coherently Driven and Superdirective Antennas**

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6 **Abstract:** Antennas are crucial elements for wireless technologies, communications and power  
7 transfer across the entire spectrum of electromagnetic waves, including radio, microwaves, THz and  
8 optics. In this paper, we review our recent achievements in two promising areas: coherently  
9 enhanced wireless power transfer (WPT) and superdirective dielectric antennas. We show that the  
10 concept of coherently enhanced WPT allows improvement of the antenna receiving efficiency by  
11 coherent excitation of the outcoupling waveguide with a backward propagating guided mode with  
12 a specific amplitude and phase. Antennas with the superdirectivity effect can increase the WPT  
13 systems performance in another way, through tailoring of radiation diagram via engineering  
14 antenna multipoles excitation and interference of their radiation. We demonstrate a way to  
15 achieving the superdirectivity effect via higher-order multipoles excitation in a subwavelength  
16 high-index spherical dielectric resonator supporting electric and magnetic Mie multipoles. Thus,  
17 both types of antenna discussed here possess a coherent nature and can be used in modern  
18 intelligent antenna systems.

19 **Keywords:** Antennas, Coherently Driven Antennas, Superdirective Antennas  
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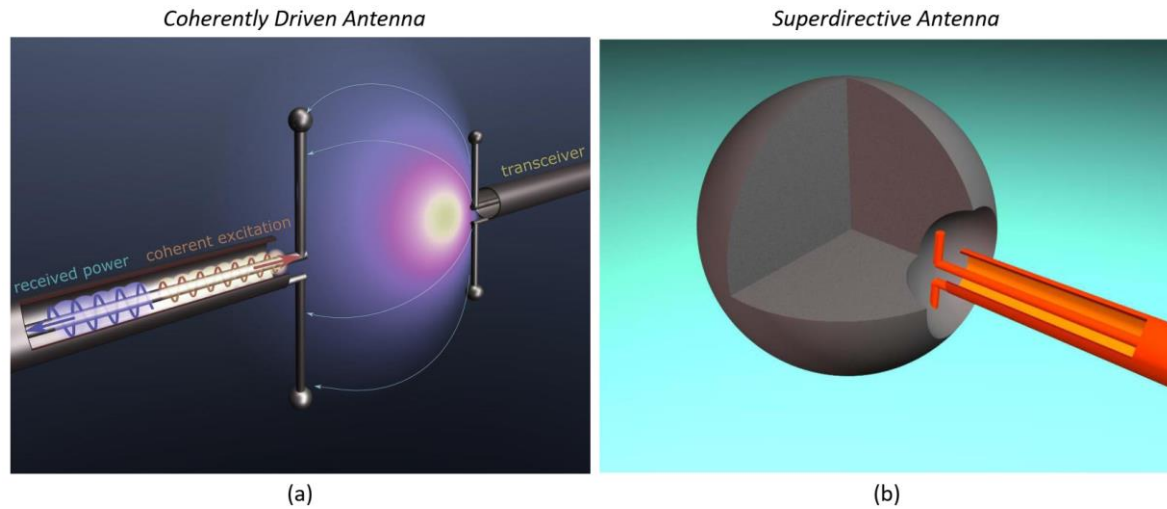
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### 21 **1. Introduction**

22 An antenna is a key element for many vital wireless technologies, including communications  
23 and power transfer[1]. First antennas emerged at the same time with the discovery of electromagnetic  
24 waves by H. Hertz in 1888 and since then have been developing with human civilization often being  
25 a catalyst for its development. A plethora of antennas have been invented in the radio and microwave  
26 frequency ranges, including microstrip antennas[2], reflector antennas[1,3], dielectric antennas[4,5]  
27 to mention just a few. More recently, so-called nanoantennas or antennas operating in the optical  
28 frequency range have been invented, which become irreplaceable elements for quantum optics and  
29 communications on a chip[6–10].

30 While wireless communications are rather established, the *wireless power transfer* (WPT),  
31 proposed at the beginning of the 20th century by N. Tesla is experiencing a rebirth. It was caused by  
32 an experimental demonstration in Ref.[11] that the WPT efficiency can be drastically enhanced when  
33 the power is transferred via resonant coupling. In that paper, wireless energy transfer between two  
34 metallic coils over the distance of 2 m with 45% efficiency in the kHz range via strongly coupled

35 magnetic resonances was demonstrated[11]. These fascinating results have given rise to many novel  
 36 technologies, include implanted devices, electric vehicles, and consumer electronics[12]. Since then,  
 37 significant research efforts have been devoted to exploring the ways to achieve as high WPT  
 38 efficiency as possible[13], whereas the main part of them has been concentrated on optimizing of  
 39 electromagnetic resonators' geometry, surrounding material, their relative arrangement.



40 (a) Operation principle of coherently enhanced wireless power transfer: a transmitting  
 41 antenna (right) radiates radio waves to the receiving antenna (left) driven by an additional wave sent  
 42 from the circuit to the antenna (red wave). (b) Superdirective dielectric antenna: subwavelength  
 43 spherical dielectric resonator (shape can be different) with a small notch on its surface fed by a short  
 44 dipole source placed into the notch. Inhomogeneous near field of the dipole source excites high-order  
 45 multipoles in the resonator which radiate coherently and can form the superdirectivity regime.  
 46

47 On another hand, designing of highly directive antennas is a significant problem in the theory  
 48 of antennas, which has been worrying researchers and engineers for a long time. The importance of  
 49 highly directional antennas for wireless interconnections and power transfer can be understood  
 50 considering the Friis equation (for matched antennas)

$$51 \quad P_{\text{rec}} = \eta_r \eta_t D_r D_t |\mathbf{a}_r \cdot \mathbf{a}_t^*| \frac{\lambda_0^2}{(4\pi d)^2} P_{\text{tr}}, \quad (1)$$

52 which says that the transmitted power ( $P_{\text{rec}}$ ) scales as product of receiver and transmitter antennas'  
 53 directivities ( $D_r D_t$ ) and also depends on their radiation efficiencies ( $\eta_r$  and  $\eta_t$ ), operation  
 54 wavelength ( $\lambda_0$ ), separation distance ( $d$ ), relative polarization ( $\mathbf{a}_r \cdot \mathbf{a}_t^*$ ), and total transmitted power  
 55 ( $P_{\text{tr}}$ ). Hence, if directivities of the both antennas are the same ( $D$ ) then the transmitted power scales  
 56 as  $D^2$ , and hence can be significantly enlarged by using more directive antennas. There are various  
 57 approaches to achieve a sufficiently high directivity for practical purposes relying on large

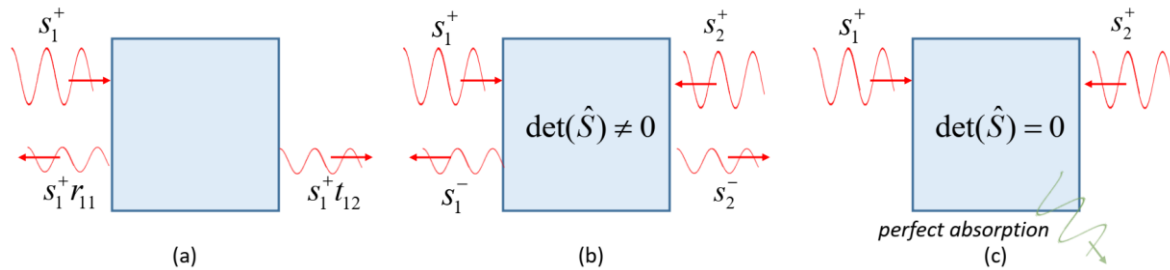
58 geometrical dimensions, including Yagi-Uda antenna, lens antenna, leaky-wave antenna, and others.  
59 All these approaches rely on the radiated (received) wave propagation in the antenna structure and,  
60 in result, these antennas have a large size ( $l \gg \lambda_0$ ) in at least one direction. However, many  
61 application areas require an antenna to be directive and compact ( $l < \lambda_0$ ) at the same time. Such  
62 antennas whose size is smaller than the operation wavelength in all three directions and directivity  
63 much larger than the directivity of a short dipole antenna (1.5) are called *superdirective antennas*[14–  
64 21]. Superdirective antenna operation relies on creating rapidly spatially oscillating currents in a  
65 subwavelength area, which leads to excitation of higher multipoles[14,21]. In result, the antenna  
66 becomes directive despite its subwavelength volume. Usually the superdirectivity regime is achieved  
67 in arrays of short dipoles, which of them fed with a specific amplitude and phase. This approach  
68 appeared to be unstable and energy-consuming and, in result, did not find a widespread practical  
69 application.

70 In this paper, we present our recent achievements in *coherently enhanced wireless power transfer*  
71 and *superdirective antennas*[21–23]. Figure 1(a) demonstrates schematically the operation principle of  
72 coherently enhanced wireless power transfer. Any WPT system consists of at least two antennas:  
73 transmitting and receiving ones. The transmitting antenna (right) radiates radio waves toward  
74 receiving antenna (left), which receives some power. We have demonstrated that the same receiving  
75 antenna driven by an additional wave (red wave) can either receive more power that it did without  
76 the coherent excitation or become more stable to changes in environment or relative arrangement.  
77 Next section is dedicated to such coherently enhanced WPT antennas. Figure 1(b) demonstrates the  
78 superdirective dielectric antenna consisting of a subwavelength spherical dielectric resonator (shape  
79 can be different) with a small notch on its surface and fed by a short dipole source placed in the notch.  
80 Inhomogeneous near field of the dipole source excites high-order multipoles in the resonator which  
81 radiate coherently and can form the superdirectivity regime. This approach is discussed in the Section  
82 3.

## 83 2. Coherently enhanced wireless power transfer

84 The concept of coherently enhanced WPT is based on the fundamental property of the wave  
85 nature of electromagnetic field - interference. Only recently it has been clearly emphasized that  
86 electromagnetic processes such as absorption and scattering may be effectively controlled via  
87 coherent spatial and temporal shaping of the incident electromagnetic field. For example, a coherent  
88 perfect absorber (CPA) is a linear electromagnetic system in which perfect absorption of radiation is  
89 achieved with two or more incident waves, creating constructive interference inside an absorbing  
90 structure. In this section, we show that the same principle that underlies the operation of CPAs can  
91 be employed to improve the efficiency of WPT systems[22]. More specific, by coherent excitation of

92 the receiving antenna [see Figure 1(a)] with an auxiliary signal, tuned in sync with the impinging  
 93 signal from the transmitting antenna, it is possible to enhance the efficiency and robustness of the  
 94 WPT system. This additional signal improves energy transfer through constructive interference with  
 95 the impinging wave, compensating any imbalance in the antenna coupling without having to modify  
 96 the load.



97  
 98 **Figure 2.** System with two channels, for example, dielectric resonator slab. (a) Excitation of this  
 99 system from one channel by a wave  $s_1^+$  results in partial transmission ( $s_1^+ t_{12}$ ) and reflection ( $s_1^+ r_{11}$ ).  
 100 (b) Excitation by two waves from both channels ( $s_1^+$  and  $s_2^+$ ) result in outgoing waves in both  
 101 channels ( $s_1^-$  and  $s_2^-$ ) if  $\det(\hat{S}) \neq 0$ . (c) Tailoring the structure geometry (thickness) and adding  
 102 loss causes CPA regime ( $\det(\hat{S}) = 0$ ), when all incoming energy gets dissipated. Changing of  
 103 relative amplitude or phase of  $s_1^+$  and  $s_2^+$  causes changing of mutual absorption and transmission.

104 Since the concept of coherently enhanced WPT is relative to CPA effect, first we discuss the main  
 105 aspects of coherent perfect absorption. Coherent perfect absorption generalizes the concept of  
 106 ordinary perfect absorbers to systems with two or more excitation channels[24–28]. This phenomenon  
 107 can be illustrated by consideration of a two-port planar structure, Figure 2. Excitation of this resonator  
 108 from one channel by a wave  $s_1^+$  results in partial transmission ( $s_1^+ t_{12}$ ) and reflection ( $s_1^+ r_{11}$ ).

109 However, excitation by two waves from both channels ( $s_1^+$  and  $s_2^+$ ) result in outgoing waves in both  
 110 channels ( $s_1^-$  and  $s_2^-$ ) and can be described by the  $\hat{S}$  matrix, which links waves in inputs and

111 outputs,  $\begin{pmatrix} s_1^- \\ s_2^- \end{pmatrix} = \hat{S} \begin{pmatrix} s_1^+ \\ s_2^+ \end{pmatrix}$ , where  $\hat{S} = \begin{pmatrix} r_{11} & t_{12} \\ t_{21} & r_{22} \end{pmatrix}$ . Here  $s_i^+$  and  $s_i^-$  respectively denote the input

112 and output wave amplitudes in the  $i$ -th channel;  $r_{ii}$  are the reflection coefficients in each port, and  
 113  $t_{ij}$  are the transmission coefficients. This  $\hat{S}$  matrix has the following eigenvalues

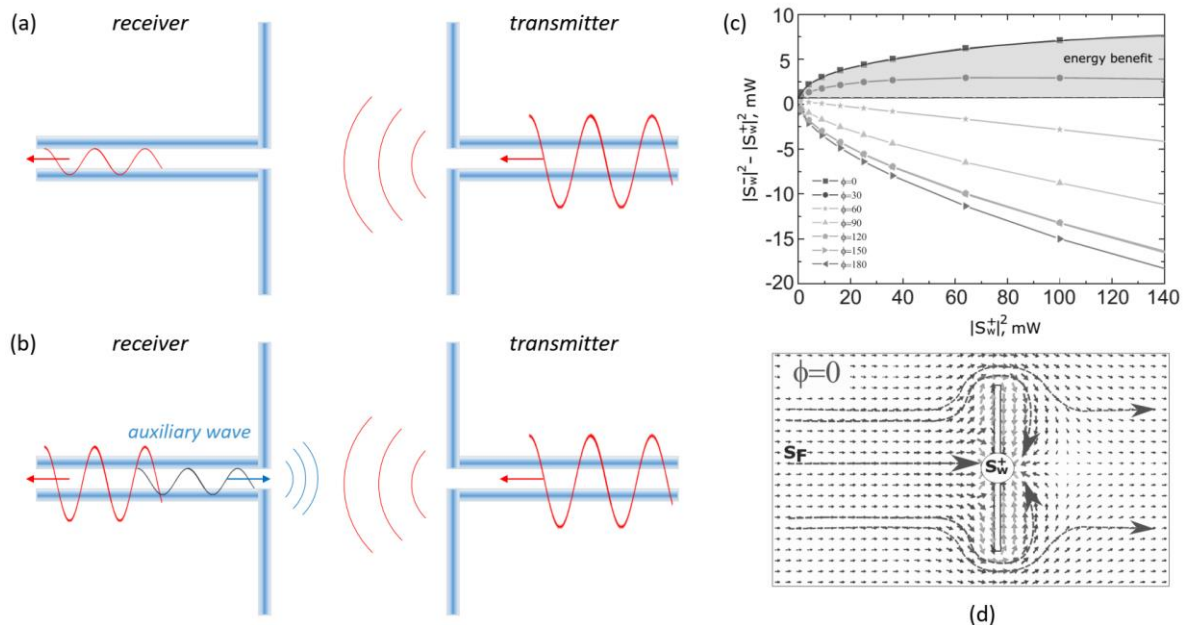
114  $d_{1,2} = \frac{1}{2} \left( r_{11} + r_{22} \pm \sqrt{r_{11}^2 - 2r_{11}r_{22} + r_{22}^2 + 4t_{12}t_{21}} \right)$  and eigenvectors

115  $\mathbf{s}_{1,2} = \left\{ 1, r_{11} - r_{22} \pm \sqrt{r_{11}^2 - 2r_{11}r_{22} + r_{22}^2 + 4t_{12}t_{21}} / 2t_{21} \right\}$ . If  $\hat{S}$  -matrix is constrained by optical

116 reciprocity[29], then  $t_{12} = t_{21} = t$ . If in addition the two-port structure is symmetric under mirror  
 117 reflection,  $r_{11} = r_{22} = r$ , the corresponding eigenvalues and eigenvectors become  $d_{1,2} = r \pm t$  and  
 118  $\mathbf{s}_{1,2} = \{1, \pm 1\}$  representing *symmetric and antisymmetric inputs* of equal intensity.

119 Next, if the system is lossless, all zeros of  $d_{1,2} = r \pm t$  (i.e.,  $\det(\hat{S}) = 0$ ) are located in the upper  
 120 complex frequency plane and we always have some outgoing waves in both channels ( $s_1^-$  and  $s_2^-$ ).  
 121 However, the situation drastically changes when we add a certain amount of material losses. In this  
 122 case, it is possible to get the S-matrix zero ( $\det(\hat{S}) = 0$ ) at the real axis, when  $r \pm t = 0$ . Then  
 123 excitation of the resonator by the corresponding CPA eigenmode  $\mathbf{s}_{1,2} = \{1, \pm 1\}$  will lead to complete  
 124 absorption, i.e. CPA effect, Figure 2(c). It turns out that the absorption in CPA is highly sensitive to  
 125 variations of the excitation conditions due to its coherent nature. Changing of relative amplitude or  
 126 phase of  $s_1^+$  and  $s_2^+$  causes changing of absorption, which can be characterized by joint absorption

127 [30]  $A \equiv 1 - \frac{|s_1^-|^2 + |s_2^-|^2}{|s_1^+|^2 + |s_2^+|^2}$ . Here,  $|s_1^+|^2 + |s_2^+|^2$  is proportional to the total input intensity, and  
 128  $|s_1^-|^2 + |s_2^-|^2$  to the total output intensity;  $A = 1$  in the CPA regime[24,26,30].



129  
 130 **Figure 3.** (a) Traditional WPT system consisting of transmitter antenna (right) and receiver antenna  
 131 (left). (b) Coherently enhanced wireless power transfer system with the receiver antenna driven  
 132 coherently by an auxiliary wave (shown in blue). (c) Net extracted energy as a function of the  
 133 auxiliary signal intensity for different relative phase values. Filled area indicates the region of energy  
 134 benefit, where the coherently assisted energy balance exceeds that for  $s_w^+ = 0$ . (d) Poynting vector  
 135 distribution around the antenna with the auxiliary signal  $s_w^+$  and relative phase of 0 deg.

136 In our recent paper we have shown that the same principle that underlies the operation of a CPA  
 137 can be employed for improving the efficiency of WPT[22]. More specific, we have shown that there  
 138 is a possibility to improve the receiving efficiency of an antenna by coherent excitation of the  
 139 outcoupling waveguide with a backward propagating guided mode with specific properties, Figure  
 140 1(a). Figures 3(a) and (b) demonstrate this idea. Figure 3(a) shows the traditional WPT system  
 141 consisting of transmitter antenna (right) and receiver antenna (left). Next, we assume that the receiver  
 142 antenna is driven coherently by an auxiliary wave (shown in blue), Figure 3(b). We have shown that  
 143 the analysis of this system can be performed in the framework of the temporal coupled mode theory  
 144 (TCMT)[31–33]. This analysis assumes that the receiving antenna has a single mode with the real  
 145 eigenfrequency  $\omega_0$  and the mode amplitude  $a$ , normalized such that  $|a|^2$  is the energy of the  
 146 mode. The dipole antenna couples to the waveguide and free space radiation with the coupling  
 147 constants  $|\kappa\rangle = \{\kappa_w, \kappa_f\}$ . The excited antenna mode can decay to both channels with the total  
 148 dumping rate,  $1/\tau = 1/\tau_w + 1/\tau_f$ . Next, the antenna mode is excited by is the vector of input  
 149 amplitudes  $|s_+\rangle$  consisting of the field of the transceiver  $s_f$  and the auxiliary field of coherent  
 150 excitation  $s_w^+$ . The results this analysis gives us the following formula for the amplitude of the  
 151 received field ( $s_w^-$ , extracted field)[22]

$$152 \quad s_w^- = -s_w^+ + \kappa_w \frac{\kappa_w s_w^+ + \kappa_f s_f}{i(\omega - \omega_0) + 1/\tau}. \quad (2)$$

153 The typical results that this equation yields for the net extracted energy are shown in Figure 3(c). The  
 154 area where the net extracted energy exceeds that for  $s_w^+ = 0$  (filled area) corresponds to the energy  
 155 benefit. In these regimes, the antenna revives more energy from the transceiver than it does without  
 156 coherent excitation even after subtraction of this additional energy. Interesting that in such a regime  
 157 of positive net extracted energy the Poynting vector distribution around the antenna demonstrates  
 158 many flow lines ending by the dipole antenna, Figure 3(d). Otherwise, when the relative phase  
 159 between  $s_w^+$  and  $s_f$  is 180 deg., there are a few of Poynting vector lines flowing into the antenna  
 160 that operates in the radiation regime (radiates more than receives).

161 In a practical WPT device, the amplitude and phase of the additional coherent signal can be  
 162 controlled in real time to adjust the antenna as a function of changes in the environment, temperature  
 163 changes in the load, and distance of the transmitter.

### 164 3. Superdirective antennas



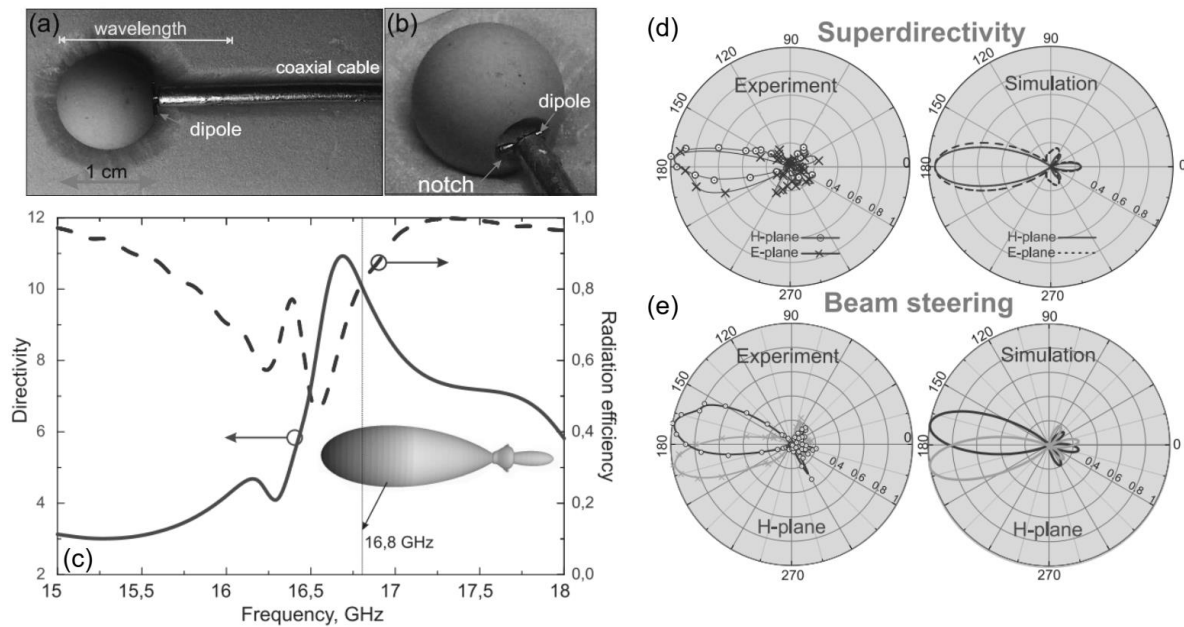
165 Usually, small antennas like dipole electric or magnetic source possess weak directivity close to that  
166 of a point dipole ( $\sim 1.5$ )[34]. Here, the directivity parameter is defines as  $D = 4\pi \max[p(\theta, \varphi)] / P_{\text{tot}}$   
167 , where  $P_{\text{tot}}$  is the total radiated power and  $p(\theta, \varphi)$  is the power density of radiation in the  
168 direction  $(\theta, \varphi)$ . Typical directive antennas such as Yagi-Uda, lens antennas, refractive antennas  
169 rely on wave propagation/refraction and hence have dimensions much larger than the operation  
170 wavelength,  $l \gg \lambda_0$ . However, as was mentioned in the introduction, there are many application  
171 areas where electrically small antennas ( $l < \lambda_0$ ) yet very directive are required. This caused intensive  
172 research on electrically small radiating systems whose directivity exceeds significantly that of a  
173 dipole. Such radiative systems were called *superdirective*[14–21]. Superdirectivity regime is of a  
174 particular interest for space communications and radio astronomy. Moreover, achieving high  
175 radiation directivity is also important for actively studied optical nanoantennas[35–37].

176 First attempts to achieve superdirectivity regime have been made in the microwave frequency  
177 range where it has been demonstrated that antenna arrays can work in this regime in a very narrow  
178 frequency range for a sophisticated system of phase shifters[14,16,34,38–40]. Physically, operation  
179 principles of superdirective antennas rely on creating rapidly spatially oscillating currents in a  
180 subwavelength area, which leads to excitation of higher multipoles[14,21]. Coherent interference of  
181 field radiated from all phased dipole or, more general, multipoles leads to the formation of a narrow  
182 lobe of power pattern. The antenna becomes directive despite its subwavelength volume. Later, the  
183 appearance of the concept of metamaterials[41] – artificially engineered materials with  
184 electromagnetic properties at will – caused a new research interest in electrically small directive  
185 antennas[42–45].

186 Recent studies on high-index dielectric antennas made of resonators possessing both resonant  
187 electric and magnetic responses made it feasible to invent new efficient antennas in  
188 microwave[21,46,47] and optics[20,48–52]. It was shown that such subwavelength high-index  
189 (usually  $n \geq 4$ ) dielectric spherical resonators possess electric and magnetic Mie  
190 resonances[21,46,47,53,54], which can be utilized for antenna designing. Although, so-called  
191 dielectric resonator antennas are known for decades[55,56], in these studies the critical role of the  
192 magnetic Mie resonance has been uncovered first in optics[8,57–59] and later in microwaves[46] and  
193 THz due to the scalability of the Maxwell's equations.

194 A very well-known system of electric and magnetic dipoles radiating coherently with the same  
195 phase is called Huygens source[1,60,61]. In result, this source consisting of two multipoles exhibit  
196 nearly twice higher directivity than that of a single electric dipole. It has been demonstrated, both in  
197 optics and microwaves, that a simple high-index dielectric resonator (spherical or cylindrical) can  
198 operate in the Huygens source regime[48,54]. This leads us to the idea that if many multipoles are

199 excited in a system, the directivity of this system can be much greater. In the same time, the size of  
 200 this dielectric antenna may be sufficiently subwavelength simply because of its high refractive index.  
 201 This idea has been suggested and realized in our recent works[10,20,21].



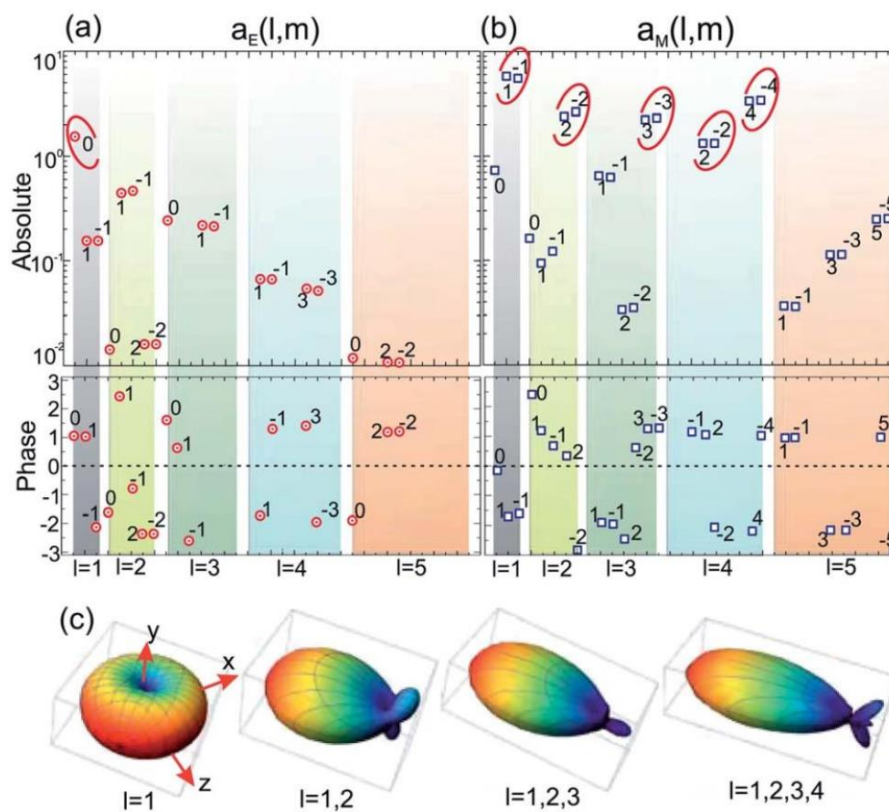
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203 **Figure 4.** Superdirective dielectric antenna. (a),(b) Photos of the realized antenna. The antenna  
 204 consists of a dielectric spherical resonator of size  $\sim \lambda_0 / 2.5$  with a small notch and a short dipole  
 205 source placed inside the notch (b). The dipole fed by a coaxial cable. (c) Results of numerical  
 206 simulations of the antenna's directivity and radiation efficiency. (d), (e) Results of experimental  
 207 realization in an anechoic chamber of the superdirectivity effect (d) and the beam steering effect (e)  
 208 and their comparison with numerical simulation results.

209 Our approach to superdirective emission is based on making a small notch on a subwavelength  
 210 high-refractive index spherical dielectric resonator supporting electric and magnetic Mie resonances  
 211 and placing a dipole source inside this notch, Figures 4(a),(b). This notch breaks the symmetry and  
 212 increases the contribution of higher-order multipoles[21,23]. The realized antenna consists of a  
 213 dielectric spherical resonator of size  $\sim \lambda_0 / 2.5$  and operates at  $\sim 17$  GHz. As a material of the  
 214 resonator was MgO-TiO<sub>2</sub> ceramic spheres characterized at microwaves by permittivity of  $\sim 16$  and  
 215 small loss tangent. This spherical resonator made of MgO-TiO<sub>2</sub> with the radius 5 mm exhibits lowest  
 216 Mie resonance at the frequency of  $\sim 14$  GHz. The radius of the semi-spherical notch was 2 mm, the  
 217 length of the dipole source was 1.5 mm. In principle, the notched antenna can be formed from a  
 218 variety of semiconductor materials and can have various shapes – spherical, ellipsoidal, cubic or  
 219 conical. The dipole fed by a coaxial cable, Figure 4(a). Figure 4(c) shows the results of numerical  
 220 simulations of the antenna's directivity and radiation efficiency. We observe a large directivity ( $\sim 11$ )



221 for such small antenna. This analysis and the analysis of the effective antenna aperture clearly  
 222 demonstrates that the antenna operates in the superdirectivity regime. The power pattern of this  
 223 antenna has been measured in an anechoic chamber and the results are presented in Figures 4(d), (e)  
 224 on the left panel. We see that these results are in a good agreement with the numerical simulations.  
 225 Remarkably, this antenna exhibits so-called beam steering effect preserving the superdirectivity  
 226 regime when the dipole source experiences a small shift to the left/right, Figure 4(e).



227

228 **Figure 5.** Magnitude (upper row) and phase (lower row) of the most contributing (a) electric and (b)  
 229 magnetic multipole moments. Multipole coefficients providing the largest contribution to the  
 230 antenna directivity are highlighted by red circles. (c) Dependence of the radiation pattern of dielectric  
 231 superdirective antenna on the number of considered multipoles (the dipole source is oriented along  
 232 the z-axis).

233 Coherent nature of the superdirectivity effect in this dielectric antenna can be revealed by  
 234 multipole decomposition technique, Figure 5, which consists in following. First, we use numerical  
 235 simulation to calculate the internal electric and magnetic field distributions. Next, knowing these  
 236 fields one can calculate the distributions of densities of (bound) carriers  $\rho = 1/(4\pi)\text{div}\mathbf{E}$  and  
 237 currents  $\mathbf{j} = c/(4\pi)(\text{rot}\mathbf{H} + ik\mathbf{E})$ . Finally, using these sources one can calculate the spherical  
 238 harmonic electric and magnetic coefficients  $a_E(l, m)$  and  $a_M(l, m)$ , which characterize the  
 239 electrical and magnetic multipole moments[62]:

$$240 \quad a_E(l, m) = \frac{4\pi k^2}{i\sqrt{l(l+1)}} \int Y_{lm}^* \left( \rho \frac{\partial}{\partial r} [rj_l(kr)] + \frac{ik}{c} (\mathbf{r} \cdot \mathbf{j}) j_l(kr) \right) d^3r, \quad (3.1)$$

$$241 \quad a_M(l, m) = \frac{4\pi k^2}{i\sqrt{l(l+1)}} \int Y_{lm}^* \operatorname{div} \left( \frac{\mathbf{r} \times \mathbf{j}}{c} \right) j_l(kr) d^3r, \quad (3.2)$$

242 where  $Y_{lm}$  are spherical harmonics of order  $(l > 0, 0 \leq m \leq l)$ ,  $c$  is the light velocity, and  
 243  $k = 2\pi / \lambda_0$ .

244 Figures 5(a),(b) demonstrate the results of the multipole decomposition of the antenna in the  
 245 superdirective regime. Here, the magnitude (upper row) and phase (lower row) of the most  
 246 contributing electric (a) and magnetic (b) multipole moments are presented. Multipole coefficients  
 247 providing the largest contribution to the antenna directivity are highlighted by red circles. We see  
 248 that despite the small antenna size, it supports multipoles of high order, including electric and  
 249 magnetic dipole ( $l = 1$ ), magnetic quadrupole ( $l = 2$ ), octupole ( $l = 3$ ) and so on.

250 We have also demonstrated that this superdirectivity regime is accompanied by a significant  
 251 increase of the effective near field zone of the antenna compared to that of a point dipole for which  
 252 the near zone radius is  $\sim \lambda_0 / 2\pi$  [23].

253 The results of multipole decomposition allow us to retrieve the power pattern on the antenna  
 254 and study the role of each multipole to the superdirectivity regime formation. Figure 5(c)  
 255 demonstrates the dependence of the radiation pattern of the dielectric superdirective antenna on the  
 256 number of considered multipoles (the dipole source is oriented along the z-axis). This result shows  
 257 that the directivity grows as higher order multipole terms are added to the response. We see that the  
 258 coherent emission from many high-order multipoles is crucial for the superdirectivity formation.

## 259 5. Conclusions

260 Both types of antenna discussed here, coherently driven and superdirective all-dielectric can be  
 261 utilized in modern intelligent antenna systems in not only radio and microwaves but also in optics.  
 262 This paper has been dedicated to the analysis of our recent works in this area. We have shown that  
 263 the concept of coherently enhanced WPT allows improvement of the antenna receiving efficiency by  
 264 coherent excitation of the outcoupling waveguide with a backward propagating guided mode with  
 265 a specific amplitude and phase. Antennas with the superdirectivity effect can increase the WPT  
 266 systems performance in another way, through tailoring of radiation diagram via engineering antenna  
 267 multipoles excitation and interference of their radiation. We have demonstrated a way to achieving  
 268 the superdirectivity effect via higher-order multipoles excitation in a subwavelength high-index  
 269 spherical dielectric resonator supporting electric and magnetic Mie multipoles.

270 **Conflicts of Interest:** The authors declare no conflict of interest.

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