Article

Model Predictive Control for a Recover Water Process

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- Abstract: The goal of this contribution is an application of the Linear General Model Predictive
- 2 Control (LGMPC). In this paper, stability of the LGMPC is proven by means of a demonstration
- of a Theorem stating a sufficient and constructive c ondition. This condition can be applied for
- 4 calculating the weight matrices of the cost function in the optimisation problem in LGMPC. Lower
- bounds conditions are found for one of these matrices and then a system with saturation is taken into
- consideration. The conditions could be interpreted and discussions through physical aspects. The
- obtained results were tested by means of computer simulations and an example with a recover water
- process is considered.
- Keywords: Linear Model Predictive Control; Process Control; Stability

1. Introduction

Model predictive control approach is used for improving the tracking of a desired trajectory. Utilizing the linear prediction algorithm for improving the tracking performances of an adaptive controller can be considered a good idea. Using a prediction structure an improvement of the dynamic performances is awaited., so MPC is used in drives control applications [1,2]. MPC is an optimal control approach being able to deal effectively with constraints and multi variable processes in industries. For 15 its being advantageable MPC has been widely used in automotive and process control communities [3]. Solving optimization problem on-line however limits the MPC applications by slow dynamic systems [3]. Application of MPC to mechatronic systems for servo design attracts attention of many scientists because of the continuous development of microprocessor technology. Mechatronic systems 19 such as electrical motor control [4], two stage actuation system control, machine tool chattering control 20 [5] have shown promising results. There is also a fast development of various advanced techniques integrated with MPC for performance improvement [6]. The used sampling frequency considered in the research literature may be still too slow for general mechatronic systems. Moreover, simulations show, that the existence of modelling error makes the steady-state error being obvious. These both points, in particular, during real-time implementation may influence the performance of mechatronic 25 systems. Calculation of the solution of the MPC in an off-line in an explicit way can be seen, for 26 instance [7], [8]. MPC has been also applied for piezoelectric actuators, see [9]. Finding conditions on the stability is one of the interesting issues in optimisation. The goal of this paper is to find lower bounds of a matrix characterising the cost function to be able to guarantee the stability of the optimal 29 solution. Results, obtained by means of PI controller in [10] are extended in this contribution. Results 30 obtained using optimal algorithms in [11] are organised in the following way. The Section 2 is dedicated 31 to a system recovering water. This contribution shows only the "difference case" for the MPC, although a similar Theorem was proven in [12]. In Section 4 a property of the LGMPC is proved in case without and with saturation. In Section 6 the system shown in Section 2 is taken as a simulated example and the proposed control technique is applied. Conclusions close the paper.

 G_p : Model predictive control state matrix

 \mathbf{F}_{2p} : Model predictive control input matrix

 \mathbf{F}_{1p} : Model predictive control "dealta" input matrix

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2. Mathematical model of the system

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The main nomenclature $m_{in}(t)$: input mass flow (kg/sec) $m_o(t)$: output mass flow (kg/sec) m(t): mass (Kg) $\frac{dm(t)}{dt}$: mass flow (kg/sec) p(t): pressure inside the evaporator (Pa) *T*: Temperature (K) V: Volume (l) p_i : initial pressure (Pa) p_d : desired pressure (Pa) R_g : vapor constant A_w : anti-windup signal K_b : weight factor for the anti-windup action $V_{\mathcal{L}}$: a Lyapunov function A_k : discrete system matrix \mathbf{B}_k : discrete input matrix \mathbf{H}_k : output matrix

In Fig. 1 a schematic representation of the considered system is shown. The system basically consists of the following three elements: a boiler, a compressor and an evaporator. In the boiler the waste water remains in the middle and lower part of it and in the upper part a vapor chamber is to be considered in order to host the water vapor which is generated by heating the waste water. At the beginning the mixture (waste water) is heated by a resistor system until the water vapor is present in the vapor chamber. At this time the heating is switched off and the compressor is switched on. The compressor should provide for a high pressure in the evaporator and in the meantime reduce the pressure in the vapor chamber through and a mass flow. This mass flow $(m_0(t))$ after the condensed phase represents cleaned water in output of the evaporator and it is stilled. More in depth, in the evaporator, the vapor is condensed giving up heat to the waste water. To guarantee that new waste water can enter into the boiler from the upper part of it, the compressor provides a low pressure in the vapor chamber. In this sense, the compressor works as a controller in this process and its output is represented by the mass flow $m_0(t)$ with the constraint that $m_0(t) > 0$. The compressor is equipped with an asynchronous motor which is controlled in frequency using an inverter piloted by a PWM signal which converts the output of the LMPC controller in frequency. In input to LMPC the pressure error is provided. More details of the control scheme will be given in the chapter devoted to the simulation. In any case, the dynamics of the asynchronous motor together with the inverter and all rest of the converters, considered all together, are much faster than the dynamics of the controlled process and in this analysis are not taken into account. Short description of the drain: At the beginning the pressure in the container equals the ambient pressure (circa 1.013 bar). By heating, the pressure increases slowly, as the water begins to evaporate. New water is added by a constant control with floats. The condensed water vapor from the "internal heat exchanger" is now free of foreign matter. For controller design purposes the dynamical model of the system must be considered. As already

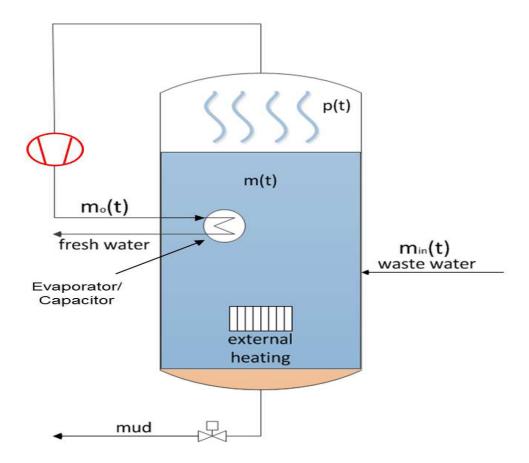


Figure 1. Boiler system

explained, considering that the control process starts after the heating phase and more precisely when the water vapor becomes to be present in the boiler, the following equations can be considered:

$$\frac{\frac{dm(t)}{dt} = m_i(t) - m_o(t)}{\frac{dp(t)}{dt} = \frac{dm(t)}{dt} \frac{R_g T}{V},}$$
(1)

in which $m_i(t)$ is a stepwise positive constant function.

Considering the Forward Euler discretisation with sampling time T_s , the following expression is obtained:

$$m(k+1) = T_s(m_i(k) - m_o(k)) + m(k) p(k+1) = (m_i(k) - m_o(k)) \frac{R_g T}{V} + p(k),$$
(2)

and thus

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$$\underbrace{\begin{bmatrix} m(k+1) \\ p(k+1) \end{bmatrix}}_{\hat{\mathbf{z}}(k+1)} = \underbrace{\begin{bmatrix} T_s & 0 \\ 0 & 1 \end{bmatrix}}_{\mathbf{A}_k} \underbrace{\begin{bmatrix} m(k) \\ p(k) \end{bmatrix}}_{\hat{\mathbf{z}}(k)} + \underbrace{\begin{bmatrix} T_s \\ T_s R_g T \\ V \end{bmatrix}}_{\mathbf{B}_k} \underbrace{(m_i(k) - m_0(k))}_{u_{mpc(t)}} \tag{3}$$

41 3. Model predictive control

In the model approach just two samples are considered:

$$\hat{\mathbf{z}}(k+1/k) = \mathbf{H}_k \mathbf{A}_k \mathbf{z}(k/k) + \mathbf{H}_k \mathbf{B}_k u_{mnc}(k). \tag{4}$$

If $\Delta u_{mpc}(k) = u_{mpc}(k) - u_{mpc}(k-1)$, then $u_{mpc}(k) = \Delta u_{mpc}(k) + u_{mpc}(k-1)$ and

$$\hat{y}_h(k+1) = \mathbf{H}_k \mathbf{A}_k \hat{\mathbf{z}}(k/k) + \mathbf{H}_k \mathbf{B}_k (\Delta u_{mpc}(k) + u_{mpc}(k-1)). \tag{5}$$

where matrix \mathbf{H}_k selects the second state variable and thus $y_h(t)$ represents the pressure.

$$\hat{y}(k+2) = \mathbf{H}_k \mathbf{A}_k^2 \hat{\mathbf{z}}(k/k) + \mathbf{H}_k \mathbf{A}_k \mathbf{B}_k (\Delta u_{mpc}(k) + u_{mpc}(k-1)) + \mathbf{H}_k \mathbf{B}_k (\Delta u_{mpc}(k+1) + u_{mpc}(k)).$$
 (6)

It is straightforward to show that the following vectorial expression holds:

$$\hat{\mathbf{Y}}_h(k) = \mathbf{G}_p \mathbf{x}(k) + \mathbf{F}_{1p} \Delta \mathbf{U}_{mpc}(k) + \mathbf{F}_{2p} u_{mpc}(k-1), \tag{7}$$

where

$$\hat{\mathbf{Y}}_h(k) = \begin{bmatrix} \hat{y}_h(k+1) \\ \hat{y}_h(k+2) \end{bmatrix}, \qquad \Delta \mathbf{U}_{mpc}(k) = \begin{bmatrix} \Delta u_{mpc}(k) \\ \Delta u_{mpc}(k+1), \end{bmatrix}$$
(8)

and matrices G_p , F_{1p} and F_{2p} are given by:

$$\mathbf{F}_{1p} = \begin{bmatrix} \mathbf{H}_k \mathbf{B}_k & \mathbf{0} \\ \mathbf{H}_k (\mathbf{A}_k \mathbf{B}_k + \mathbf{B}_k) & \mathbf{H}_k \mathbf{B}_k \end{bmatrix}, \qquad \mathbf{G}_p = \begin{bmatrix} \mathbf{H}_k \mathbf{A}_k \\ \mathbf{H}_k \mathbf{A}_k^2 \end{bmatrix}, \qquad \mathbf{F}_{2p} = \begin{bmatrix} \mathbf{H}_k \mathbf{B}_k \\ \mathbf{H}_k (\mathbf{A}_k \mathbf{B}_k + \mathbf{B}_k) \end{bmatrix}. \tag{9}$$

If the following performance criterion is assumed,

$$J = \frac{1}{2} \sum_{j=1}^{N} \left(y_d(k+j) - \hat{y}(k+j) \right)^{\mathrm{T}} \mathbf{Q}_p \left(y_d(k+j) \right) - \hat{y}(k+j) \right) + \sum_{j=1}^{N} \left(\Delta u_{mpc}(k+j) \right)^{\mathrm{T}} \mathbf{R}_p \left(\Delta u_{mpc}(k+j) \right), \tag{10}$$

where $y_d(k+j)$, j=1,2,...,N is the pressure reference profile and N the prediction horizon, and \mathbf{Q}_p and \mathbf{R}_p are non-negative definite matrices. Index (16) can be written as

$$J = \frac{1}{2} \hat{\mathbf{Y}}_h^T(k) \mathbf{Q}_p \hat{\mathbf{Y}}_h(k) + \frac{1}{2} \Delta \mathbf{U}_{mpc}^T(k) \mathbf{R}_p \Delta \mathbf{U}_{mpc}(k), \tag{11}$$

then the solution minimizing performance index (11) may then be obtained by solving

$$\frac{\partial J}{\partial \Delta \mathbf{U}_{mnc}} = 0. \tag{12}$$

A direct off-line computation may be obtained explicitly as

$$\Delta \mathbf{U}_{mpc} = (\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_p \mathbf{F}_{1p} + \mathbf{R}_p)^{-1} \Big(\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_p \big(\mathbf{Y}_{d_p}(k) - \mathbf{G}_p \mathbf{z}(k) - \mathbf{F}_{2p} u_{mpc}(k-1) \big) \Big), \tag{13}$$

where $\mathbf{Y}_{d_v}(k)$ is the desired output column vector. For further details see [13].

4. A stability sufficient constructive condition in GMPC

Theorem 1. Let us take the discrete SISO linear system into consideration:

$$\mathbf{z}(k+1) = \mathbf{A}_k \mathbf{z}(k) + \mathbf{B}_k u_{mpc}(k), \tag{14}$$

$$y(k) = \mathbf{H}_k \mathbf{z}(k), \tag{15}$$

obtained by a discretisation of a linear continuous system using a sampling time which equals T_s . $u_{mpc}(k)$ represents the first element of the vector of the optimal solution as calculated in [13] for the GMPC considering the following cost function:

$$J = \frac{1}{2} \sum_{j=1}^{N} \left(y_d(k+j) - \hat{y}(k+j) \right)^{\mathrm{T}} \mathbf{Q}_p \left(y_d(k+j) - \hat{y}(k+j) \right) + \sum_{j=1}^{N} \left(\Delta u_{mpc}(k+j-1) \right)^{\mathrm{T}} \cdot \mathbf{R}_p \Delta u_{mpc}(k+j-1), \quad (16)$$

where $y_d(k+j)$, $j=1,2,\ldots,N$ is the position reference trajectory and N is the prediction horizon, and \mathbf{Q}_p and \mathbf{R}_p are non-negative definite matrices. Furthermore, the solution minimizing performance index (16) may be obtained by solving:

$$\frac{\partial J}{\partial \Delta \mathbf{U}_{mpc}} = 0. \tag{17}$$

It is known for instance from [13] that the optimal solution is:

$$u_{mpc}(k) = (\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_p \mathbf{F}_{1p} + \mathbf{R}_p)^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_p \Big(\mathbf{Y}_{d_p}(k) - \mathbf{G}_p \mathbf{z}(k) - \mathbf{F}_{2p} u_{mpc}(k-1) \Big), \tag{18}$$

- where $\mathbf{Y}_{d_n}(k)$ and $\mathbf{Y}_p(k)$ are the desired output column vector and the measured or observed output vector.
- Matrices Q_p and R_p are diagonal and positively defined. Under the technical hypotheses that $Q_p = I$, and
- **H**_k^T**H** = **I** are under the assumption
 - i) $r_{(1,1)} >> T_s^2$, where $r_{(1,1)}$ represents the first diagonal element of matrix \mathbf{R}_p . then \forall $r_{(1,1)}$ such that:

$$r_{(1,1)} > \mathbf{B}_k \mathbf{B}_k^{\mathrm{T}} \frac{\|\mathbf{A}_k\|_2}{1 - \|\mathbf{A}_k\|_2},$$
 (19)

where $\|\mathbf{A}_k\|_2$ represents the maximal eigenvalue of matrix $\sqrt{\mathbf{A}^T\mathbf{A}}$ and

$$\|\mathbf{A}_{k} - \mathbf{B}_{k} \Big((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{G}_{p} \Big) \|_{2} < \|\mathbf{A}_{k} + \mathbf{B}_{k} \Big((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{G}_{p} \Big) \|_{2}, \quad (20)$$

then the system (14) is asymptotically stable.

Proof 1. For sake of brevity just one prediction step is considered, then:

$$\mathbf{F}_{1p} = \left[\mathbf{H}_k \mathbf{B}_k \right], \tag{21}$$

$$\mathbf{F}_{2p} = \left[\mathbf{H}_k \mathbf{B}_k \right], \tag{22}$$

$$\mathbf{G}_p = \left[\mathbf{H}_k \mathbf{A}_k \right]. \tag{23}$$

Combining Eq. (14) with (18), this expression is got:

$$\mathbf{z}(k+1) = \mathbf{A}_k \mathbf{z}(k) + \mathbf{B}_k \Big((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_p \mathbf{F}_{1p} + \mathbf{R}_p)^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_p \Big(\mathbf{Y}_{d_p}(k) - \mathbf{G}_p \mathbf{z}(k) - \mathbf{F}_{2p} u_{mpc}(k-1) \Big) \Big), \quad (24)$$

which is equivalent to write:

$$\mathbf{z}(k+1) = \left(\mathbf{A}_{k} - \mathbf{B}_{k} \left((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{G}_{p} \right) \right) \mathbf{z}(k) +$$

$$\mathbf{B}_{k} \left((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \left(\mathbf{Y}_{d_{p}}(k) \right) \right)$$

$$- \mathbf{B}_{k} \left((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \left(\mathbf{F}_{2p}(k) \right) \right).$$
 (25)

If

$$r_{(1,1)} > \mathbf{B}_k \mathbf{B}_k^{\mathrm{T}} \frac{\|\mathbf{A}_k\|_2}{1 - \|\mathbf{A}_k\|_2},$$
 (26)

considering that scalar $r_{(1,1)} > 0$ and scalar $\mathbf{B}_k \mathbf{B}_k^{\mathrm{T}} > 0$, then:

$$0 < \|\mathbf{A}_k\|_2 + r_{(1,1)}^{-1} \mathbf{B}_k \mathbf{B}_k^{\mathrm{T}} \|\mathbf{A}_k\|_2 < 1.$$
 (27)

Recalling that $\mathbf{H}_k^{\mathrm{T}}\mathbf{H} = \mathbf{I}$

$$0 < \|\mathbf{A}_k\|_2 + r_{(1,1)}^{-1} \mathbf{B}_k \mathbf{B}_k^{\mathrm{T}} \mathbf{H}_k^{\mathrm{T}} \mathbf{H} \|\mathbf{A}_k\|_2 < 1$$
 (28)

and thus

$$0 < \|\mathbf{A}_k\|_2 + r_{(1,1)}^{-1} \|\mathbf{B}_k \mathbf{B}_k^{\mathrm{T}} \mathbf{H}_k^{\mathrm{T}} \mathbf{H} \mathbf{A}_k \|_2 < 1.$$
 (29)

Being matrix \mathbf{F}_{1p} defined as in (21) and \mathbf{G}_{1p} defined in (23), it is known that matrix \mathbf{B}_k is proportional T_s , then considering that $\mathbf{R}_p = r_{(1,1)}$, choosing a suitable $r_{(1,1)} >> T_s$ and considering that $\mathbf{Q}_p = \mathbf{I}$ (technical hypothesis), the following condition can be derived:

$$0 < \|\mathbf{A}_{k}\|_{2} + \left\| \left(\mathbf{B}_{k} \left((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{G}_{p} \right) \right\|_{2} < 1.$$
 (30)

Considering the norm properties and condition (20), then:

$$0 < \|\mathbf{A}_{k} - \mathbf{B}_{k} \Big((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{G}_{p} \Big) \|_{2}$$

$$< \|\mathbf{A}_{k} + \mathbf{B}_{k} \Big((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{G}_{p} \Big) \|_{2}$$

$$< \|\mathbf{A}_{k} \|_{2} + \|\mathbf{B}_{k} \Big((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{G}_{p} \Big) \|_{2} < 1. \quad (31)$$

To conclude

$$0 < \|\mathbf{A}_{k}\| - \left\|\mathbf{B}_{k}\left((\mathbf{F}_{1p}^{\mathsf{T}}\mathbf{Q}_{p}\mathbf{F}_{1p} + \mathbf{R}_{p})^{-1}\mathbf{F}_{1p}^{\mathsf{T}}\mathbf{Q}_{p}\mathbf{G}_{p}\right)\right\|_{2} < 1.$$
 (32)

The constraint in (26) states a plausible condition on the controller. Stability represents the necessary condition of the optimality of a controlled system. In order to interpret the result let us observe a mechanical system including a mass-spring system in which it is known that the eigenvalue can variate in the whole real and complex domain as a function of the mass which states the system inertia. If mass $m \to \infty$, then, because of the discretisation and according to the Landau notation, $\left(\left\|\lambda_{max}\left(\mathbf{A}_k\right)-1\right\|\right)\to 0 \text{ with } \mathcal{O}\left(\left|\lambda_{max}\left(\mathbf{A}_k\right)-1\right|\right)=\mathcal{O}(\frac{1}{m}). \text{ In the meantime } \mathcal{O}(\mathbf{B}_k\mathbf{B}_k^{\mathrm{T}})=\mathcal{O}(\frac{1}{m^2}).$ For a very slow system, according to (26), $r_{(1,1)}\to 0$, parameter $r_{(1,1)}$ is present in the denominator function of the optimal solution in (18) and in this case, small values of $r_{(1,1)}$ are devoted to speed up the system. If mass $m\to 0$, then, because of the discretisation, $\mathcal{O}\left(\left|\lambda_{max}\left(\mathbf{A}_k\right)\right|\right)=\mathcal{O}\left(\frac{1}{m}\right)$, but in the meantime $\mathbf{B}_k\mathbf{B}_k^{\mathrm{T}}\to\infty$ with $\mathcal{O}(\mathbf{B}_k\mathbf{B}_k^{\mathrm{T}})=\mathcal{O}(\frac{1}{m^2})$. For a very fast system $r_{(1,1)}\to\infty$, parameter $r_{(1,1)}$ is

devoted to slow down the system. So we can conclude that a highly inertial system needs relatively small values of $r_{(1,1)}$ to be optimised and stabilised. If the inertia is small, then the system needs larger values of $r_{(1,1)}$ to be optimised and stabilised. In fact, very fast systems can have very high abrupt changes in the input signals and in the cost function the input factor needs to be reduced to find an optimality.

5. The saturation case

Proposition 1. Let us take the following discrete SISO linear system as before into consideration:

$$\mathbf{z}(k+1) = \mathbf{A}_k \mathbf{z}(k) + \mathbf{B}_k u_{mpc}(k), \tag{33}$$

$$y(k) = \mathbf{H}_k \mathbf{z}(k), \tag{34}$$

and

$$|u_{mpc}(k)| \le U_{max} \ \forall k, \tag{35}$$

then (33) with the input saturation defined in (35) being asymptotically stable and its input avoiding the constraint if condition (26) holds together with the input limitation. The following condition summarizes the result:

$$r_{(1,1)} > \max \left\{ \frac{\mathbf{B}_{k} \mathbf{B}_{k}^{\mathrm{T}} \|\mathbf{Y}_{d_{p}}(k)\|_{2}}{U_{max}} + \frac{\|\mathbf{B}_{k} \mathbf{F}_{1p} \mathbf{Q}_{p} \mathbf{F}_{2p}\|_{2}}{U_{max}} |u_{mpc}(k-2)|, \ \mathbf{B}_{k} \mathbf{B}_{k}^{\mathrm{T}} \frac{\|\mathbf{A}_{k}\|_{2}}{1 - \|\mathbf{A}_{k}\|_{2}} \right\}.$$
(36)

Proof 2. The demonstration is straightforward just considering that:

$$|u_{mpc}(k)| < |u_{mpc}(k)| < U_{max} \ \forall k$$

and thus it is enough that the following condition holds:

$$\left\| \mathbf{B}_{k} \left((\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \left(\mathbf{Y}_{d_{p}}(k) \right) \right) - \mathbf{B}_{k} (\mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \mathbf{F}_{1p} + \mathbf{R}_{p})^{-1} \mathbf{F}_{1p} \mathbf{Q}_{p} \mathbf{F}_{2p} u_{mpc}(k-1) \right\|_{2} \le U_{max}.$$
(37)

In fact, using similar considerations as before, the following expression is obtained:

$$\left\| \mathbf{B}_{k} r_{11}^{-1} \mathbf{F}_{1p}^{\mathsf{T}} \mathbf{Q}_{p} \left(\mathbf{Y}_{d_{p}}(k) \right) \right\|_{2} + \left| \mathbf{B}_{k} \mathbf{F}_{1p} \mathbf{Q}_{p} \mathbf{F}_{2p} \right| \le U_{max}, \tag{38}$$

and thus, including also the stability condition, condition (36) follows here below again:

$$r_{(1,1)} > \max \left\{ \frac{\mathbf{B}_{k} \mathbf{B}_{k}^{\mathrm{T}} \|\mathbf{Y}_{d_{p}}(k)\|_{2}}{U_{max}} + \frac{\|\mathbf{B}_{k} \mathbf{F}_{1p} \mathbf{Q}_{p} \mathbf{F}_{2p}\|_{2}}{U_{max}} |u_{mpc}(k-1)|, \ \mathbf{B}_{k} \mathbf{B}_{k}^{\mathrm{T}} \frac{\|\mathbf{A}_{k}\|_{2}}{1 - \|\mathbf{A}_{k}\|_{2}} \right\}.$$
(39)

It is possible to observe that for large values of U_{max} the condition on the input barrier in the cost function (16) with weight $r_{(1,1)}$ is not so restrictive, so larger inputs are allowed. For small values of U_{max} the input barrier limits the values of the input and no large input values are allowed.

80 6. Simulation results

Concerning the simulation results, it is to be clarified that function $m_i(t)$ is a stepwise constant function with $m_i(t) = 0.086$ (kg/sec.) or $m_i(t) = 0$ and in the simulated case $m_i(t) = 0.086$ (kg/sec.) is considered. As already explained, in the simulations two cases should be distinguished: weak anti-saturating action:

$$r_{(1,1)} > \frac{\mathbf{B}_k \mathbf{B}_k^{\mathrm{T}} \|\mathbf{Y}_{d_p}(k)\|_2}{U_{max}}.$$
 (40)

and strong anti-saturating action:

$$r_{(1,1)} >> \frac{\mathbf{B}_k \mathbf{B}_k^{\mathsf{T}} \|\mathbf{Y}_{d_p}(k)\|_2}{U_{max}}.$$
 (41)

Fig. 2 shows the controlled pressure which represents the main result. In case of anti-windup action being relatively weak, the anti-windup controller needs time to reestablish the control loop. As already explained, during the windup effect the feedback control is broken so the absence of the feedback control action can clarify the long permanence of the pressure at negative values. In particular, from Fig. 3 which represents the mass flow $\frac{dm(t)}{dt} = m_{in}(t) - m_0(t)$ (kg/sec) it is visible how this function is consequent with the relation:

$$\frac{dm(t)}{dt} \approx \frac{dp(t)}{dt}. (42)$$

Moreover, considering the following equation again:

$$\frac{dm(t)}{dt} = m_{in}(t) - m_0(t), (43)$$

to activate the process a strong initial action through the mass flow $m_0(t)$ is required. Observing these two figures it is possible too see that in the case of strong anti saturating action represented in Fig. 5 the controlled system through the anti-windup scheme comes out from the saturation state very quickly with a resulting faster dynamics. The resulting faster dynamics are obtained thanks to the stronger anti-windup action which allows to reactivate the control loop very quickly. As already explained, when the saturation occurs the control loop is open and no feedback control is present.

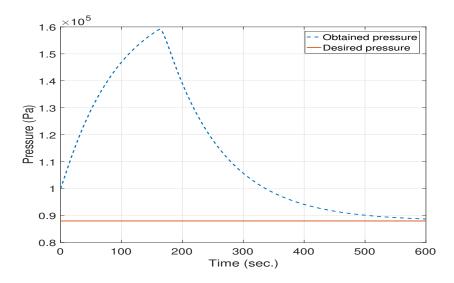


Figure 2. Desired and obtained pressure with weak anti-saturating action

7. Conclusion

One of the most important problems in the context of optimisation using LMPC is represented by conservative conditions on the stability. This paper presents a sufficient and constructive condition for the stability of an LGMPC calculating lower bounds of the element of matrix ${\bf R}$ which represent the weights of the input in a typical given cost function. A physical interpretation of the result is given on the light of some physical considerations. An illustrative example is provided in which a recover water process is taken into consideration to test the proposed results through computer simulations.

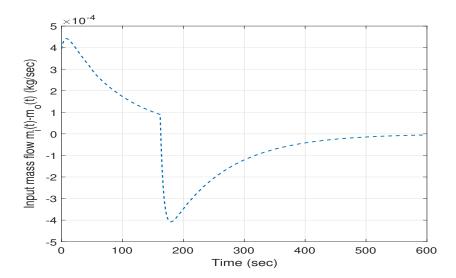


Figure 3. Mass flow $\frac{m(t)}{dt} = m_{in}(t) - m_0(t)$ (kg/sec) with weak anti-saturating action

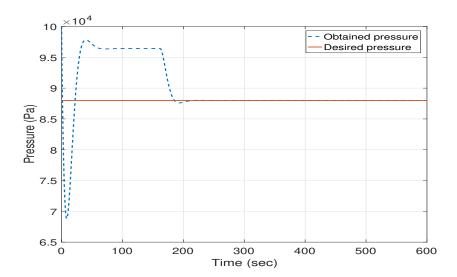


Figure 4. Desired and obtained pressure with strong anti-saturating action

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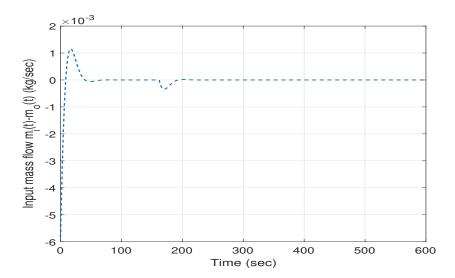


Figure 5. Mass flow $\frac{dm(t)}{dt} = m_{in}(t) - m_0(t)$ (kg/sec) with strong anti-saturating action

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