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Article

Variability Analysis of Battery EIS Measurements

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Abstract

Electrochemical Impedance Spectroscopy (EIS) is a non-destructive technique for characterizing the battery behavior for estimating the state of health (SOH). EIS provides frequency-domain information on key parameters, including solid electrolyte interphase (SEI) resistance, charge-transfer (CT) resistance, and Ohmic resistance, which are sensitive to battery degradation mechanisms. In an EIS test, a sinusoidal excitation signal is applied to the battery, and the corresponding voltage response is analyzed to extract the impedance spectrum. The reliability of SOH estimation therefore depends critically on the accurate and repeatable extraction of impedance features. This paper investigates the variability in impedance spectra arising from the state of charge (SOC), temperature, rest time, and repeated measurements under nominally identical conditions. This variability is identified as drift and represents previously underexplored variations in the impedance spectrum. To quantify these variations, this work proposes a normalized resistance-based index that captures changes in the impedance spectrum using estimated equivalent circuit model (ECM) parameters. The proposed index is applicable across battery chemistries, sizes, and operating conditions. It is evaluated using published datasets spanning different chemistries, SOC levels, and temperatures, as well as laboratory data collected from repeated EIS experiments. Results show that even at fixed SOC and temperature, repeated measurements can produce measurable bias and variance in ECM parameters. These findings highlight the importance of accounting for drift in EIS analysis and motivate uncertainty-aware battery diagnostics for practical SOH monitoring systems.

Keywords: electrochemical impedance spectroscopy (EIS); battery diagnostics; state of health (SOH); feature extraction; uncertainty quantification; equivalent circuit model; charge transfer resistance; machine learning

1. Introduction

Electrochemical Impedance Spectroscopy (EIS) is a non-invasive, frequency-domain diagnostic technique that characterizes the dynamic response of electrochemical systems by applying a small-amplitude AC perturbation and measuring the resulting voltage or current over a wide range of frequencies [1,2]. The resulting impedance spectrum acts as a unique fingerprint of the system, encapsulating diverse electrochemical and physical phenomena, such as charge transfer resistance, double-layer capacitance, diffusion impedance, and solid electrolyte interphase (SEI) behavior [3]. In lithium-ion batteries, EIS has long been used for battery characterization, equivalent circuit parameter extraction, and fault detection [4]. More recently, EIS has been recognized for its value in battery management systems (BMS) for estimating the state of health (SOH), owing to its ability to capture subtle degradation signatures that are often missed by time-domain diagnostics like voltage relaxation or incremental capacity analysis [2,4].

Several methods have been developed to extract SOH-relevant features from EIS data. A widely adopted approach is to fit the impedance spectrum to an equivalent circuit model (ECM) [5], such as a Randles circuit, which yields interpretable parameters like Ohmic resistance, SEI resistance, charge

transfer (CT) resistance, double layer (DL) capacitance, and Warburg diffusion coefficients. These parameters evolve in characteristic ways with aging and are often used as inputs to predictive models. Similarly, the distribution of relaxation times (DRT) method deconvolves the impedance spectrum into a set of relaxation processes, helping to isolate overlapping degradation mechanisms, such as SEI growth and ion transport limitations [6]. While both ECM and DRT methods provide valuable physical insights, they require full-spectrum EIS measurements and iterative numerical fitting, making them less suitable for real-time or embedded applications. To address this, alternative methods have emerged that use impedance at a single frequency or a small number of characteristic frequencies. These approaches reduce measurement time and computational overhead [7] while retaining sensitivity to key aging phenomena, making them promising for integration into practical SOH monitoring systems.

With the increasing availability of high-quality EIS datasets, machine learning (ML) and deep learning (DL) models have become integral to SOH estimation [8,9]. These models can be trained using either physically derived features (e.g., ECM or DRT parameters) or the raw impedance spectra. Traditional ML algorithms, such as support vector regression (SVR) and shallow neural networks have been used in conjunction with feature selection or optimization algorithms to map the extracted impedance features to SOH. More recently, deep learning architectures, such as convolutional neural networks (CNNs), recurrent neural networks (e.g., LSTM), and vision transformer (ViT)-based models have enabled direct learning from high-dimensional impedance data [10]. These models excel at capturing complex nonlinear dependencies between spectral patterns and battery degradation, and some have demonstrated robustness across varying temperatures, state of charge (SOC) levels, and usage profiles.

While end-to-end models that consume raw impedance spectra can offer high accuracy, they often require large training datasets and lack interpretability. In contrast, analytical features derived from ECM or DRT methods are physically meaningful and often generalize better across cells and operating conditions, even with limited training data. Moreover, reduced-feature set approaches based on carefully chosen frequencies strike a balance between accuracy, interpretability, and computational efficiency. As a result, there is growing interest in using these analytical or semi-analytical features as inputs to ML models, especially in resource-constrained or safety-critical environments [11]. These hybrid strategies, rooted in electrochemical understanding yet powered by data-driven learning, present a compelling direction for the deployment of EIS-based SOH diagnostics in future battery management systems.

While analytical features derived from EIS data offer interpretability and reduced model complexity, their usefulness is highly dependent on the accuracy and consistency with which they are extracted. Improperly or inconsistently extracted features can introduce significant uncertainty into the SOH estimation process. In Table 1, a detailed review of prior studies that have identified various sources of uncertainty in the EIS spectra is provided.

Table 1. Summary of experimental studies investigating the variability in battery impedance spectrum.

No	Factors	References	Observations about the Impedance Spectrum
1	State of Health (SOH)	[12–18]	The impedance of a battery is known to increase with battery age. The impedance rise is attributed to an increasing time constant, SEI growth, lithium-ion reduction, electrode performance decline, and capacity loss. For aged cells, an additional semicircle can be observed in the impedance spectrum at high frequencies. At lower frequencies, the real part of the impedance spectrum shifts rightward as the battery ages. In the sensitivity analysis in [15], the impact of age on EIS was ranked the highest, followed by temperature and SOC.

Table 1. Summary of experimental studies investigating the variability in battery impedance spectrum (continued).

No	Factors	References	Observations about the Impedance Spectrum
2	Temperature	[19],[12], [20], [21], [22], [13], [23], [24], [25] [14] ,[15] [26]	Both the real and imaginary parts of the battery impedance are affected by temperature [20,22,26]. Temperature increase is known to accelerate the lithium-ion transfer rate and conductivity. The Ohmic resistance dominated by the electrolyte, SEI resistance, and CT resistance increases with decreasing temperature [21,26]. By decreasing the temperature by 10 °C, the CT resistance is reported to have increased by a factor of 2.4-2.7 in [13]. The highest difference in the imaginary part of the impedance due to temperature is observed between frequency ranges of 5 kHz to 10 kHz.
3	State of Charge (SOC)	[6], [27] , [12], [28], [29], [20], [30], [21], [31], [32] [14] [33] [15],[26],[34]	EIS measurement of Li-ion batteries with low SOC level shows higher impedance value [12],[28], [29], [20]. At lower frequencies, the real part of the impedance spectrum decreases gradually as the SOC increases [29],[30], [21], [31] [26] and reaches about 90%, then slightly increases as SOC reaches 100% [6]. At 100% SOC, the impedance increases because there are fewer lithium ions at the cathode [27]. The Nyquist plot “expands” as SOC decreases [20] and “contracts” as it increases from about 10% to 90% [6]. The total resistance also decreases with increasing SOC [34].
4	Resting Time	[12], [6], [35], [36], [37], [38], [39], [29], [13].	When subsequent EIS measurements are taken within shorter rest times, the resulting Nyquist plot is observed to show more variability due to varying polarization and diffusion effects, especially in the low-frequency regions. At higher frequencies, the real part of the impedance, i.e., ohmic resistance, remains consistent at varying rest times [6]. The impedance increases as the rest time increases at lower frequencies, with the maximum variation in impedance observed at a frequency of lower than or equal to 1 Hz [38]. In [12], it was observed that a 2-hour relaxation resulted in a stable impedance measurement, subject to only certain battery chemistries.
5	Laboratory environment	[19]	Slight differences in the impedance are observed during interlaboratory EIS comparison, also called a Round Robin evaluation. Particularly, the differences of the resistance values are below 10 $\mu\Omega$ for frequencies up to 100 Hz and reach up to 200 $\mu\Omega$ at 10 kHz. The differences are attributed to variations in device position inaccuracies, cable movement, instrument drift and random thermal noise.
6	AC excitation current	[40–42]	The selection of AC amplitude during EIS is crucial for recording the precise impedance information of the battery. Small current amplitudes result in a distorted Nyquist plot due to low signal-to-noise ratio (SNR), and as the amplitude of the AC excitation current increases, the SNR improves and the Nyquist plot becomes noise-free. Any AC amplitude above the range of 5 mA to 50 mA was considered acceptable in the literature, with SNR increasing as the amplitude increases beyond these limits [41].
7	DC component of excitation current	[28]	The impedance measured by conventional EIS methods is typically modeled as a dynamic linear time-invariant (LTI) system. The application of a superimposed DC during EIS results in time-variability and nonlinearities due to the open-circuit voltage (OCV) element of the battery, which violate the underlying LTI assumptions. Thus, the measured impedance needs to be corrected to remove the non-linear OCV effects to avoid the over-estimation of battery impedance, particularly at low frequencies.

Table 1. Summary of experimental studies investigating the variability in battery impedance spectrum (continued).

No	Factors	References	Observations about the Impedance Spectrum
8	Usage pattern	[35,40]	In prior literature, the effect of prior current pattern through the battery is defined as a crucial factor affecting the battery's impedance. This is because a higher C-rate before EIS measurement results in the rapid initial voltage drop and a slower relaxation toward stabilized voltage and temperature levels. Moreover, as the C-rate increases from 2C to 8C at shorter relaxation times, the impedance is observed to decrease, especially in the intermediate-to-low frequency regions.

For instance, even well-established features such as the CT resistance or SEI resistance can vary depending on the rest period prior to measurement, as batteries often require stabilization to reflect their true impedance characteristics. Additionally, sensor noise and resolution limitations in practical EIS measurement systems can distort feature values, particularly at low frequencies where measurement sensitivity decreases [19]. Furthermore, many EIS features are inherently dependent on operating conditions, such as SOC [12,28] and temperature [20,22,26]. If these conditions are not tightly controlled during data acquisition, or at least properly accounted for during feature interpretation, the extracted values may reflect environmental variation rather than genuine battery degradation. These uncertainties not only reduce the reliability of the estimated SOH but may also degrade the performance of machine learning models trained on such features. Therefore, it is critical to understand and quantify the uncertainty associated with the EIS feature extraction. Incorporating such uncertainty information into the estimation pipeline can improve robustness and enable probabilistic SOH predictions that better reflect real-world conditions.

In this paper, we address the critical need for uncertainty quantification in EIS-based analytical features used for battery SOH estimation. We develop mathematical models to characterize the variability of commonly used features, such as charge transfer resistance, SEI resistance, and Ohmic impedance, under different operating conditions. By leveraging the available EIS datasets from the published literature, we show variations in the estimated parameters with respect to SOC, temperature, and rest duration. By systematically modeling the sources and magnitudes of variability in EIS features, this work contributes both to the reliability of SOH predictions and to the broader development of uncertainty-aware battery management algorithms.

2. Materials and Methods

2.1. EIS Variability Modeling

In this section, we introduce parameters to capture the variability in the Nyquist plots due to various factors summarized in Table 1.

Figure 1 shows the adaptive Randles equivalent circuit model (AR-ECM) of a battery. The EMF component represents the electromotive force of the battery, which is also known as the open circuit voltage (OCV). The remainder of the AR-ECM consists of the following elements: stray inductance L , ohmic resistance R_{Ω} , solid electrolytic interface (SEI) resistance R_{SEI} , SEI capacitance C_{SEI} , charge transfer (CT) resistance R_{CT} , double layer (DL) capacitance C_{DL} , and Warburg impedance Z_w that is modelled as

$$Z_w \triangleq Z_w(j\omega) = \sigma \sqrt{\frac{2}{j\omega}} \quad (1)$$

where σ is the Warburg coefficient and ω is the angular frequency.

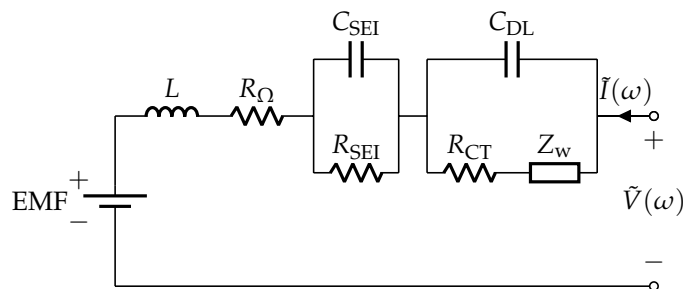


Figure 1. Adaptive Randles equivalent circuit model (AR-ECM) of a battery.

Consider a sinusoidal excitation current of frequency ω applied to the battery, and let the resulting voltage response be measured. After transforming both signals to the frequency domain, e.g., using the discrete Fourier transform (DFT), we denote the current excitation and the voltage response as

$$I(\omega) = I_{dc} + \tilde{I}(\omega) \quad (2)$$

$$V(\omega) = V_{dc} + \tilde{V}(\omega) \quad (3)$$

where $I(\omega)$ and $V(\omega)$ denote the current and voltage components consisting of the DC component (I_{dc} , V_{dc}) and AC component ($\tilde{I}(\omega)$, $\tilde{V}(\omega)$). The AC impedance of the battery at frequency ω can then be written as

$$\begin{aligned} Z(j\omega) &= \frac{\tilde{V}(\omega)}{\tilde{I}(\omega)} \\ &= j\omega L + R_{\Omega} + \frac{1}{\frac{1}{R_{SEI}} + j\omega C_{SEI}} + \frac{1}{\frac{1}{R_{CT} + Z_w(j\omega)} + j\omega C_{DL}} \\ &= j\omega L + R_{\Omega} + \frac{R_{SEI}}{1 + j\omega R_{SEI} C_{SEI}} \\ &\quad + \frac{R_{CT} + Z_w(j\omega)}{1 + j\omega (R_{CT} + Z_w(j\omega)) C_{DL}} \end{aligned} \quad (4)$$

To construct a typical battery impedance response, the AC impedance given in (4) is computed across a wide range of frequencies, typically spanning from microhertz (μHz) to megahertz (MHz).

Figure 2 presents a baseline Nyquist plot of the AR-ECM shown in Figure 1, using the parameter values in Table 2. Here, the running parameter ω runs from right to left on the curve. All parameters of the AR-ECM can be identified from the corresponding Nyquist plot [43]. The frequency range used in Figure 2 spans from 10 μHz to 10 kHz.

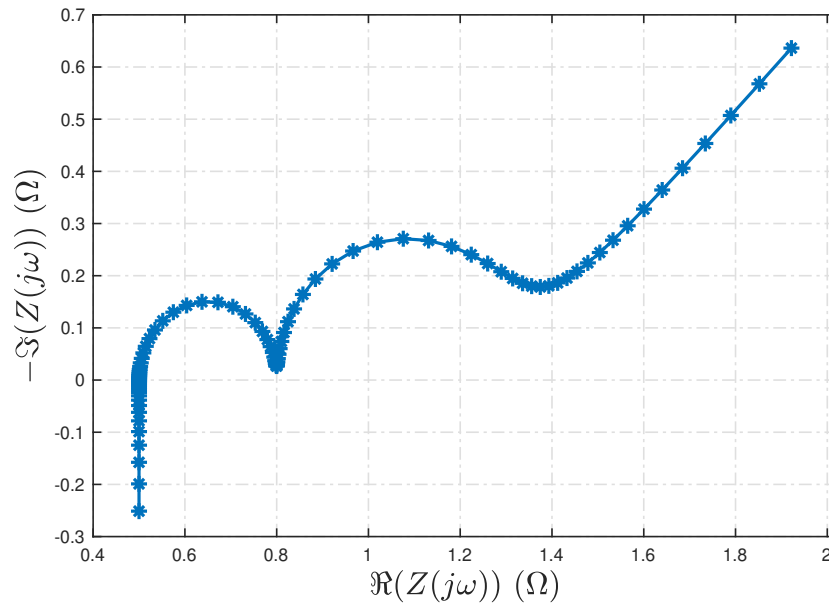


Figure 2. Baseline Nyquist plot corresponding to the AR-ECM parameters in Table 2.

Table 2. Parameters of the AR-ECM.

Parameter	Value
L	$4 \times 10^{-6}\text{H}$
R_{Ω}	0.5Ω
R_{SEI}	0.3Ω
C_{SEI}	0.2F
R_{CT}	0.5Ω
C_{DL}	100F
σ	0.005

2.1.1. Effect of Measurement Noise

It was shown in [44] that the zero-mean property of voltage and current measurement noise is preserved under transformation to the frequency domain. Leveraging this property, Figure 3 illustrates the effect of measurement noise on the battery's impedance spectrum. A distinctive feature of measurement noise is that it introduces increasing jitter in the Nyquist plot as the noise level rises. This feature is consistent with observations reported in the literature, where an increasing amount of measurement noise caused jittery features on the computed impedance spectrum [19].

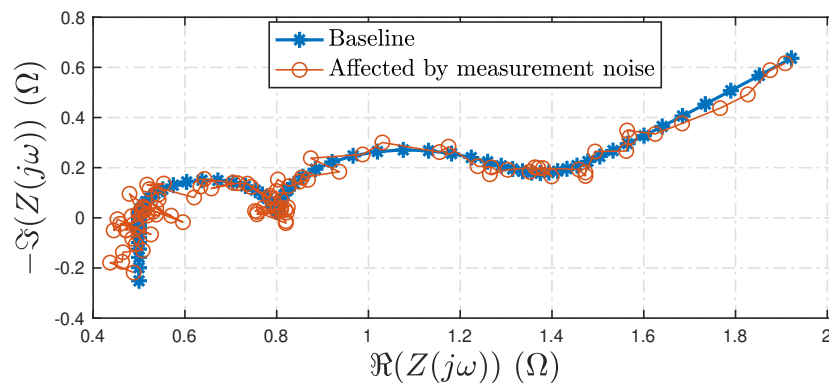


Figure 3. Effect of measurement noise in the baseline Nyquist plot in Figure 2.

2.1.2. Bias and Variation in the Resistance

Several of the observed changes and variations in the battery impedance spectrum outlined in Table 1 can be attributed to fluctuations in the Ohmic and diffusion resistance, R_d , which is defined as

$$R_d = R_{SEI} + R_{CT} \quad (5)$$

To quantify the variations in the battery impedance spectrum, the resistances are expressed relative to a reference value as

$$R_\Omega = \bar{R}_\Omega + \Delta R_\Omega \quad (6)$$

$$R_d = \bar{R}_d + \Delta R_d \quad (7)$$

where ΔR_Ω and ΔR_d represent the deviations from the reference values, \bar{R}_Ω and \bar{R}_d , respectively. Now, it is hypothesized that the deviations follow a Gaussian distribution with a mean (bias) and a standard deviation, denoted as

$$\Delta R_\Omega \sim \mathcal{N}(\mu_{R_\Omega}, \sigma_{R_\Omega}^2) \quad (8)$$

$$\Delta R_d \sim \mathcal{N}(\mu_{R_d}, \sigma_{R_d}^2) \quad (9)$$

where \mathcal{N} denotes the Gaussian distribution.

It should be noted that the Gaussian assumption is a working hypothesis and has not been rigorously validated. However, when considering cell-to-cell variability in large-scale battery systems, such an assumption may be a reasonable approximation [45–47].

The impedance of the battery consists of a real and imaginary impedance denoted as

$$Z(j\omega) = \Re(Z(j\omega)) + \Im(Z(j\omega))j \quad (10)$$

where $\Re(\cdot)$ and $\Im(\cdot)$ denote the real and imaginary parts of the impedance, respectively. Under the Gaussian assumption, the impedance is rewritten as

$$Z(j\omega) = \bar{Z}(j\omega) + \Delta Z(j\omega) \quad (11)$$

where $Z(j\omega)$ is an estimate of the impedance, $\bar{Z}(j\omega)$ is the true impedance, and $\Delta Z(j\omega)$ is the uncertainty in the impedance.

Figure 4 shows the variations in the impedance due to changes in the ohmic and diffusion resistances due to the bias effect. It can be observed from the varying bias values in Figure 4 that as the bias increases, the Nyquist plot exhibits a rightward drift, indicating an increase in the total resistance of the battery.

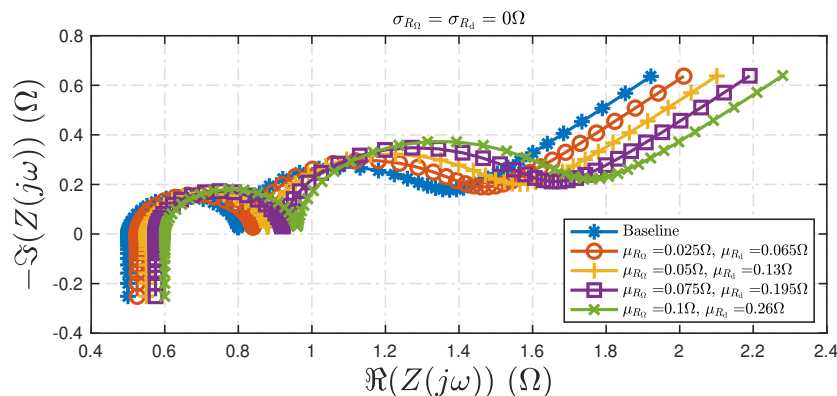


Figure 4. Variations in the impedance spectrum due to different bias values for ohmic and diffusion resistances, μ_{R_Ω} and μ_{R_d} . It is assumed that $\sigma_{R_\Omega}^2 = \sigma_{R_d}^2 = 0\Omega$.

Figure 5 shows the variations in the impedance spectrum upon changes in the variance components, $\sigma_{R_\Omega}^2$ and $\sigma_{R_d}^2$. The key difference between Figure 4 and Figure 5 lies in the nature of the variability in the impedance spectrum. In Figure 4, the shift in the Nyquist plot is systematically correlated with the bias terms. In contrast, Figure 5 reflects random variations due to the variances, resulting in non-directional expansion or contraction of the impedance spectrum with respect to a certain factor.

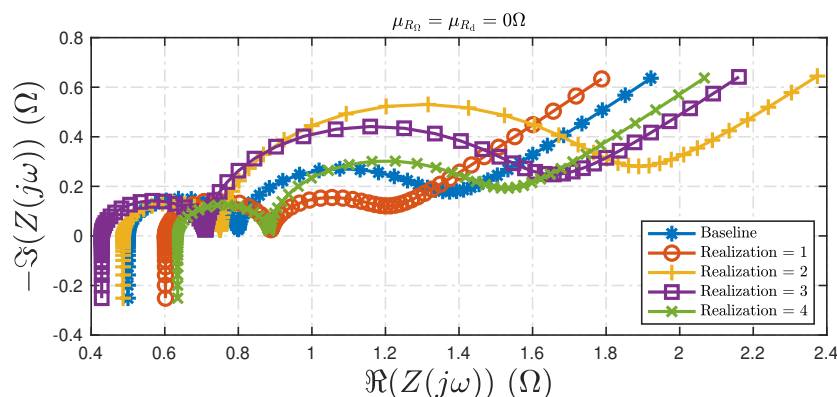


Figure 5. Variations in the impedance spectrum due to different variance values for ohmic and diffusion resistances, $\sigma_{R_\Omega} = 0.1\Omega$ and $\sigma_{R_d} = 0.26\Omega$. It is assumed that $\mu_{R_\Omega} = \mu_{R_d} = 0\Omega$.

Finally, Figure 6 shows the variations in the impedance spectrum upon changes to both the bias and variance components for the assumed Gaussian distributions in (8) and (9).

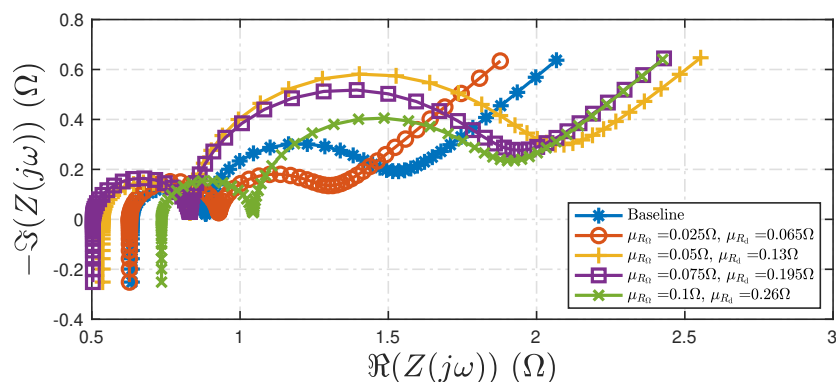


Figure 6. Variations in the impedance spectrum due to different bias and variance values for the assumed Gaussian distributions of resistances in (8) and (9). Assumed variances are $\sigma_{R_\Omega} = 0.1\Omega$ and $\sigma_{R_d} = 0.26\Omega$.

From Figure 6, it can be noticed that both the bias and variance effects result in the non-directional expansion or contraction variations in the impedance spectrum, This behavior is consistent with trends reported in the literature, where an increase in R_{CT} has been associated with certain conditions such as a decrease in state of charge (SOC) [14].

2.2. EIS Variability Metric

This section uses data from two research articles ([12] and [34]) chosen from Table 1 to demonstrate the EIS variability models introduced in Section 2.1. The variability in the impedance spectrum observed in the existing literature is compared in terms of its Ohmic and diffusion resistances.

In [34], EIS experiments are conducted on LFP and NMC batteries at varying temperature ranges from 10 to 30°C for the SOC ranges between 10 and 90%. Figure 7 summarizes the results of the Ohmic and diffusion resistance recorded under these conditions. From Figure 7((a)), it can be noticed that the Ohmic resistance is higher for the LFP batteries compared with NMC configurations at all temperatures. The diffusion resistance for the two batteries is observed to be within the range of 4 to 60 mΩ for the different experimental conditions.

More importantly, it can be observed from Figure 7 that the effect of temperature on the resistances is analogous to the impact of bias variations in the assumed distribution, as shown in Figure 4. Similarly, the effect of SOC on the resistances is comparable to the changes in the variance of the assumed distribution of the ohmic and diffusion resistances, as shown in Figure 5.

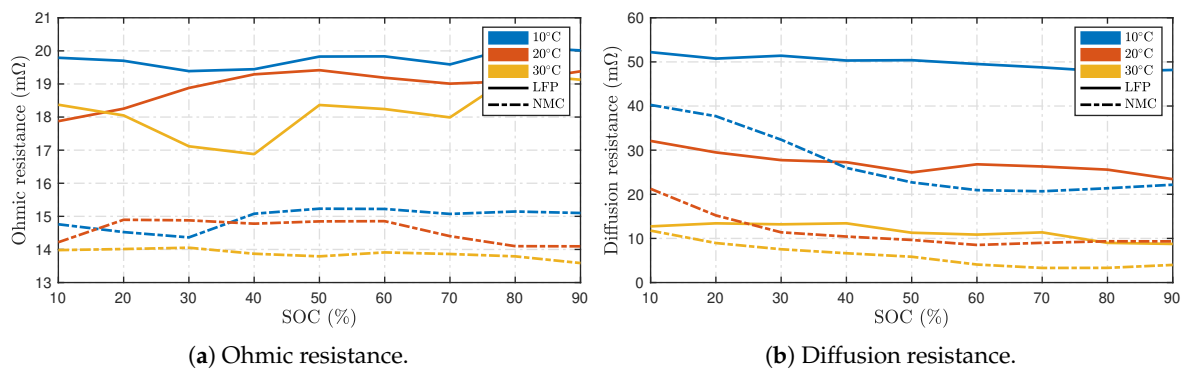


Figure 7. Estimated resistances recorded from [34].

It can be observed from the comparison in Figure 7 that the individual impedances are recorded on varying scales, and that there is no single measure by which the resistances can be compared under different experimental conditions. This is also true for batteries of different chemistries and sizes. Further, due to cell-to-cell variations, the resistance of batteries of the same chemistry, size, and identical production procedures will have slightly different resistance values due to manufacturing process variations. Hence, there is a need to define a metric that is tolerant of these variations.

To normalize the total resistance and its variation, the following metric is defined

$$d\text{-ohm} = \frac{R_d}{R_\Omega} \quad (12)$$

where $R_d = R_{SEI} + R_{CT}$ is denoted as the 'diffusion resistance' in this paper. The d-ohm metric allows assessment of battery impedance and its variability using the resistances that contribute to the impedance spectrum. The metric is computed as a ratio against the ohmic resistance to ensure that it remains a stable and definitive metric for variation in impedance spectrum — it is because the Ohmic resistance is relatively less sensitive to various factors than the diffusion resistance, as noted in Figure 7.

From Figure 8, it can be noticed that the d-ohm metric reduces as the temperature increases. The proposed metric allows for a direct comparison between LFP and NMC batteries, with LFP batteries

generally exhibiting a slightly higher d-ohm than NMC batteries across all temperatures. Finally, it can be noted that the difference in the d-ohm values is most pronounced at lower temperatures for both LFP and NMC batteries, especially at lower SOC for the NMC batteries.

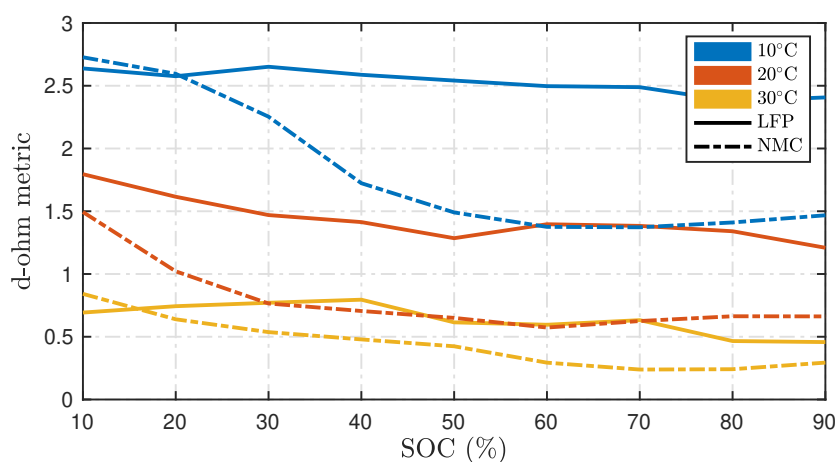


Figure 8. Proposed metric 'd-ohm' computed using (12) using estimated resistances recorded from [34].

In [12], the authors investigated three lithium-ion battery technologies, NMC, LFP and LTO at varying temperature ranges from 15 to 45°C for the SOC ranges between 20 and 80%. Figure 9 summarizes the results of the Ohmic and charge transfer (CT) resistances recorded under these conditions. From Figure 9a, it can be noticed that the Ohmic resistance is higher for the LFP than the NMC and LTO configurations. As noted also in Figure 7, the Ohmic resistance is mostly stable with minimal variation across temperatures selected in the study. The CT resistance shown in Figure 9b is observed to be higher at lower temperatures for most configurations chosen in this study. Figure 10 shows a comparison of the computed d-ohm metrics. Similar to the observations in Figure 8, it can be noticed that the d-ohm metric reduces with temperature for LFP and NMC chemistries. Contrary to the previous study, the NMC batteries exhibited a slightly higher d-ohm than the LFP batteries across all temperatures; however, the pronounced differences at lower temperatures remained the same in both studies.

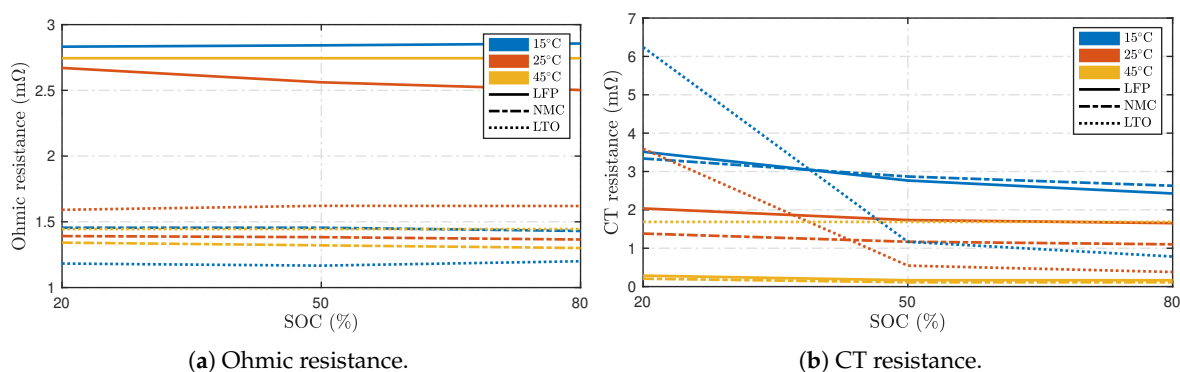


Figure 9. Change in the Nyquist plot due to variations in the Warburg coefficient.

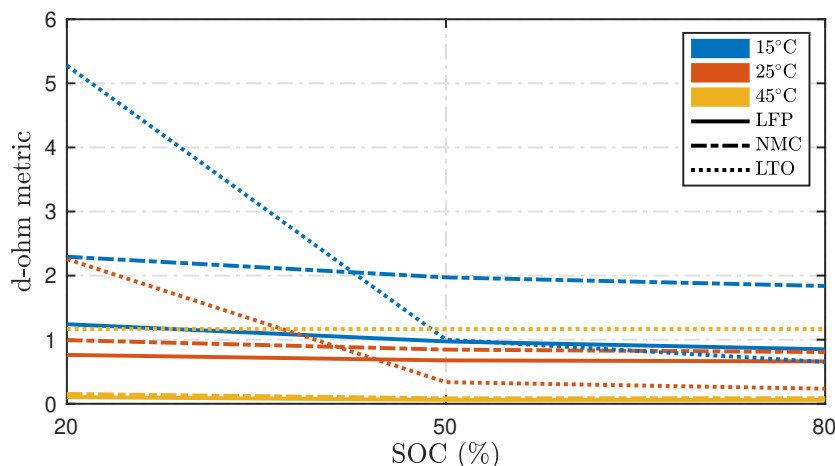


Figure 10. Proposed d-ohm metric computed using (12) using estimated resistances recorded from [34].

2.3. Experimental Setup

In this section, the experimental setup used to collect data from Li-ion batteries and the dummy cell is explained.

2.3.1. Gamry Interface 5000P

A device manufactured by Gamry Instruments called the Interface 5000P, designed for single-cell EIS testing of batteries, was used in this study, as shown in Figure 11. This instrument offers both galvanostatic and potentiostatic control modes. In both modes, the device supports currents of up to 5A, maximum voltage potential at $\pm 6V$ and allows data acquisition between 10 μHz and 20 kHz. The galvanostat mode was utilized in this study for performing the EIS of the chosen batteries.



Figure 11. Gamry Interface 5000P.

2.3.2. Dummy Cell

An AC Dummy Cell manufactured by Gamry Instruments, as shown in Figure 12 is used in this study to obtain the control measurements. This circuit can be used to simulate an RC/2RC ECM behavior with known ECM parameters given as: $R_1 = 3.01 \text{ k}\Omega$, $R_2 = 200\Omega$, $R_3 = 3.01 \text{ k}\Omega$, $C_1 = 1\mu\text{F}$, and $C_2 = 33 \text{ nF}$. The front of the dummy cell has five banana jacks (2 mm and 4 mm) to which the potentiostat cell-cable leads are connected as shown in the illustration in Figure 13. The back shows the approximation of the 2RC circuit using an RC-R-RC electrical setup with visible traces (highlighted in Figure 12).



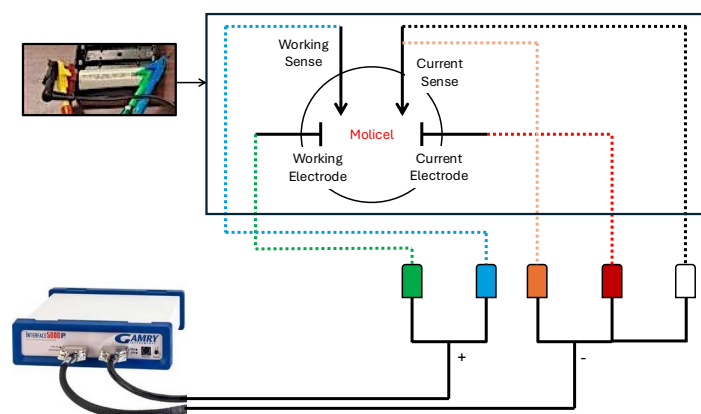
Figure 12. AC Dummy Cell.



Figure 13. Gamry Interface 5000P connections schematic to the AC Dummy Cell.

2.3.3. Battery Cell

The Molicel INR-21700-P42A lithium-ion cell, known for its high power capability and robust cycle life characteristics, is selected for the testing. This cell is manufactured in the 21700 cylindrical format and features a nickel-rich NMC (LiNiMnCoO_2) cathode paired with a graphite anode. It offers a nominal capacity of 4.2 Ah and supports continuous discharge currents of up to 45 A. The nominal voltage of the battery is 3.6V, and its typical operating voltage range is between 2.5V and 4.2V. Figure 14 summarizes how the molice is connected to the Gamry potentiostat during the experiment.



(a) Connection diagram.



(b) Physical System.

Figure 14. Gamry Interface 5000P connections to the Molicel battery.

2.3.4. Excitation Signal and Test Conditions

The EIS test was performed on both the AC Dummy Cell and the Molicel INR-21700-P42A battery using a sinusoidal current excitation signal. The excitation signal was applied across a frequency range from 500 Hz to 0.01 Hz to capture both high-frequency and low-frequency impedance characteristics. The amplitude of the sinusoidal current was set to $1 \mu\text{A}$ for the dummy cell and 100 mA for Molicel, reflecting the differences in their current-handling capabilities and ensuring that both were tested within their respective linear operating regions. A zero DC bias current with 12 complete cycles was applied in all cases to isolate the AC impedance response.

For the EIS measurements, a logarithmically spaced frequency range was employed using 10 points per decade, spanning from 0.01 Hz to 500 Hz.

2.3.5. Procedure

The experiment begins with an EIS test (Trial-1) on Day 01, conducted according to the specifications in Section 2.3.4 on the battery described in Section 2.3.3. Following this test, the battery is allowed to rest for 1 hour. After the rest period, a second EIS test (Trial-2) is performed, followed by another 1-hour rest. Next, a third EIS test (Trial-3) is conducted, after which the battery is rested for the remainder of Day 01. Subsequently, on Days 02 and 03, the three EIS trials are repeated with 1-hour resting periods between each test, marking the conclusion of the experiment. Figure 15 presents a visual summary of the experimental procedure applied to the battery.

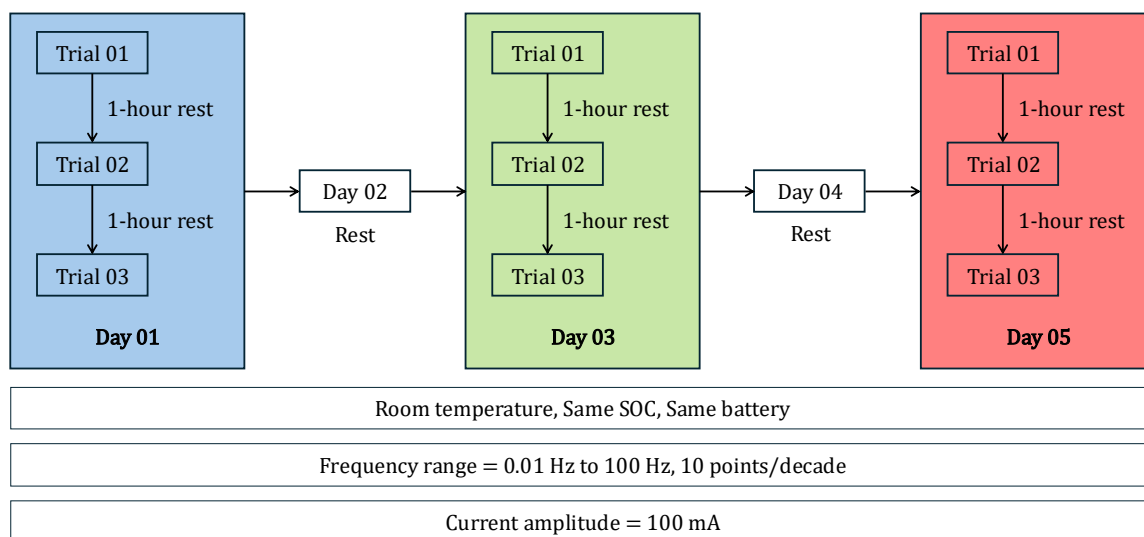


Figure 15. Experimental Setup.

The pulse frequency, real and imaginary impedance data during the three trials every day were recorded and analyzed in the next section.

3. Results

This section presents the results of the variability of the impedance spectrum based on data collected using the experimental procedure in Section 2.3.

3.1. Impedance Variability

The impedance spectrum presented in this section was recorded for the dummy cell and the battery for repeated trials.

3.1.1. AC Dummy Cell

Figure 16 shows the impedance spectrum recorded for the dummy cell for four repeated trials. From the figure, it can be noticed that all four trials appear superimposed on each other. Hence, there is minimal variation between the four trials repeated for the dummy cell.

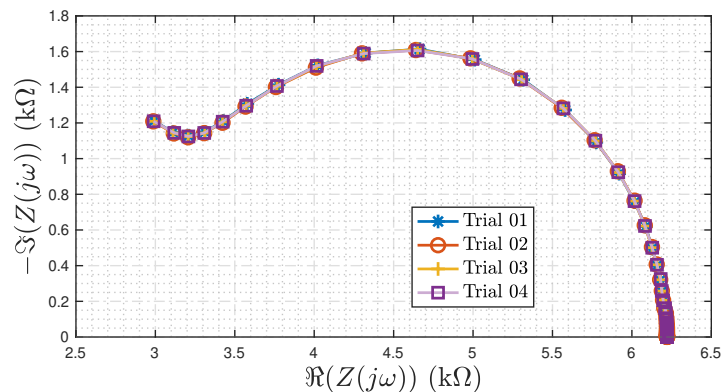


Figure 16. Impedance spectrum for four repeated trials on the Dummy Cell.

3.1.2. Battery Cell

Figure 17 shows the impedance spectrum recorded for a battery cell described in Section 2.3.3. The cell underwent three trials every day, for three days, and the impedance spectrum recorded during all trials is plotted in this Figure. Details of the experimental procedure are provided in Section 2.3.5. The progression across days is represented using color and marker variations in the figure, i.e., Day 01 is shown in blue tones, Day 02 with green tones, and Day 03 in red tones. Using the impedance data across the three days, the mean impedance is computed as

$$\bar{Z}(j\omega) = \text{mean}(\Re(Z(j\omega))) + \text{mean}(\Im(Z(j\omega)))j \quad (13)$$

In Figure 17, the mean impedance computed from the recorded data is plotted as a dotted black line.

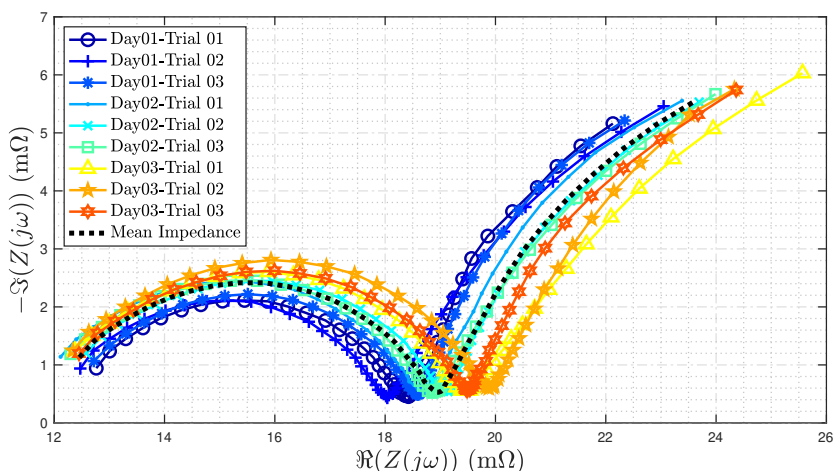


Figure 17. Impedance spectrum for three repeated trials on three different days for the battery. Day 01 is shown in blue tones, Day 02 in green tones, and Day 03 in red tones.

From Figure 17, there are several observations one can make. Firstly, it can be noticed that the real value of impedance increases with the progression of days, i.e, the impedance recorded for Day 01 is smaller than that for Day 02, followed by the highest impedance recorded during Day 03. A variation of $2\text{m}\Omega$ is observed in the recorded real impedance across the three days. This can also be noted as a maximum increase of $2\text{m}\Omega$ in real impedance from Day 01 to Day 03, solely observed during repeated EIS trials. It must be noted that within the three trials recorded on a particular day, no pattern of impedance was observed in this experiment. For instance, consider the impedance recorded during three trials on Day 01 (shown in blue tones) in Figure 17. The variation during the three trials does not follow an increasing or decreasing pattern consistent with factors affecting impedance spectrum in Table 1. In this paper, the variation observed in the impedance spectrum during repeated trials on the

same battery is defined as *drift*. The drift is observed for days 02 and 03 as well. For the trials in all three days, a drift of magnitude of 0.5mΩ was observed.

3.2. Parameter Estimation Approach

The impedance recorded for a battery can be written as a function of the AR-ECM parameters in Figure 1 using (4), i.e.,

$$Z_{\mathcal{X}}(j\omega, \mathcal{X}) = j\omega L + R_{\Omega} + \frac{R_{SEI}}{1 + j\omega R_{SEI} C_{SEI}} + \frac{R_{CT} + Z_w(j\omega)}{1 + j\omega(R_{CT} + Z_w(j\omega))C_{DL}} \quad (14)$$

where $\omega = 2\pi f$ denotes the angular frequency of the excitation signal applied across a frequency range of $f = 500$ to 0.01 Hz, and \mathcal{X} is the vector of the parameters of the AR-ECM given as

$$\mathcal{X} = \{L, R_{\Omega}, R_{SEI}, C_{SEI}, R_{CT}, C_{DL}, \sigma_w\} \quad (15)$$

The parameters \mathcal{X} can be obtained by solving the following optimization problem

$$\hat{\mathcal{X}} = \arg \min_{\mathcal{X}} \sum_{i=1}^n \left[\|\Re(Z(\omega_i)) - \Re(Z_{\mathcal{X}}(j\omega_i, \mathcal{X}))\|^2 + \|\Im(Z(\omega_i)) - \Im(Z_{\mathcal{X}}(j\omega_i, \mathcal{X}))\|^2 \right] \quad (16)$$

In this paper, the 'lsqcurvefit' function that is a part of MATLAB's Optimization Toolbox is utilized to solve the above optimization problem. The initial estimate of the parameters, along with their lower and upper limits, was defined using the maximum value of the real impedance $Z_r = \max(\Re(Z(\omega)))$, as follows,

$$\mathcal{X}_0 = [10^{-8}, 10^{-2}, Z_r/2, 10, Z_r/2, 10, 10^{-5}] \quad (17)$$

$$\mathcal{X}_{lower} = [10^{-18}, 10^{-10}, 10^{-10}, 10^{-10}, 10^{-10}, 10^{-10}, 10^{-10}] \quad (18)$$

$$\mathcal{X}_{upper} = [10^{-4}, Z_r, Z_r, 10^4, Z_r, 10^4, 1] \quad (19)$$

The limits are chosen to be smaller, more realistic values to ensure that the estimate converges. The optimization was terminated when the change in objective function, $|Z_{\mathcal{X}}(j\omega, \mathcal{X}^{k+1}) - Z_{\mathcal{X}}(j\omega, \mathcal{X}^k)|$ and step size in parameters, $|\mathcal{X}^{k+1} - \mathcal{X}^k|$ were below e^{-20} .

3.3. Estimated Parameter Analysis

Using the optimization approach described in Section 3.2, the parameters for the AR-ECM are estimated for the battery that underwent three trials every day for three days. Using the estimated parameters, the impedance spectrum is reconstructed using (4) and plotted in Figure 18 for all trials showcased in Figure 17. Using the estimated impedance, the mean impedance across the three days is computed and is also plotted in a dotted black line in Figure 18.

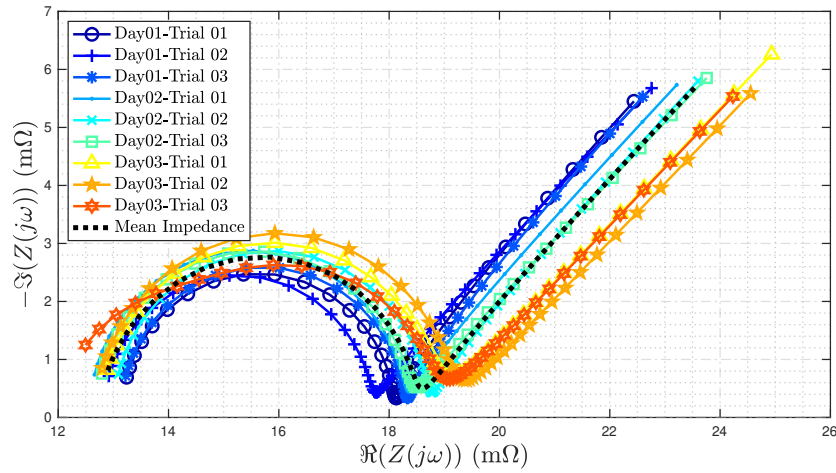
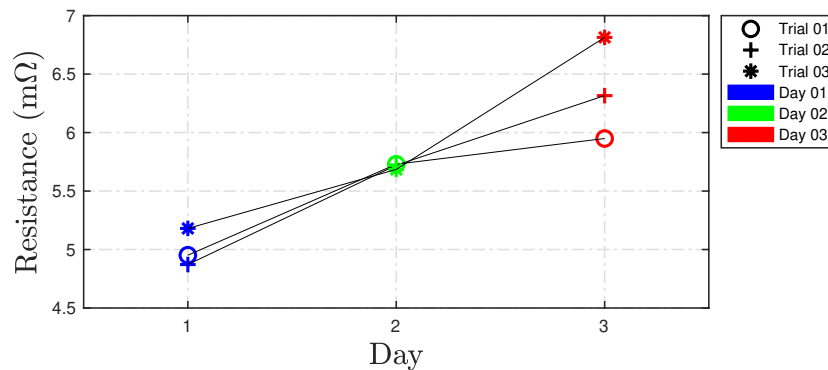
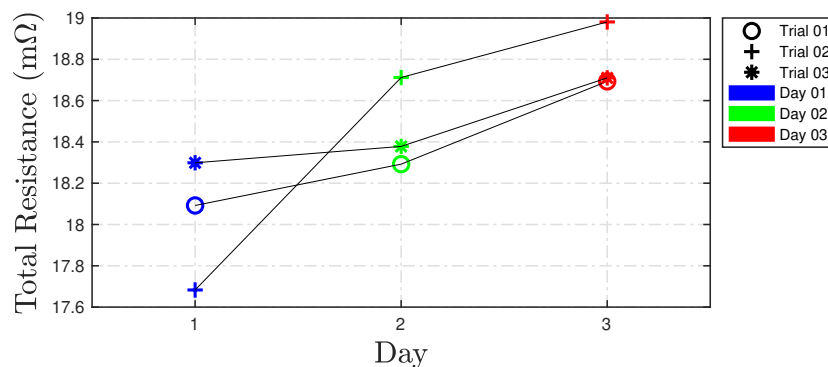


Figure 18. Impedance spectrum from estimated parameters for three repeated trials on three different days for the battery. See the corresponding measured data in Figure 17.

Now, using the estimates of R_{SEI} and R_{CT} resistances, the diffusion resistance is computed using (5) and plotted in Figure 19a. Due to the minimal variation of the individual R_{SEI} and R_{CT} resistances during Days 1 and 2, the diffusion resistance also exhibits similar patterns. Further, the diffusion resistance was observed to increase with the progression of days, i.e., the diffusion resistance recorded for Day 01 is smaller than that for Day 02, followed by the highest impedance recorded during Day 03. From Figure 19b, the variance in the diffusion resistance across the three days is observed to be 11.11%.



(a) Diffusion Resistance defined in (5).



(b) Total Resistance defined in (20).

Figure 19. Estimated resistance parameters.

Similarly, using the estimates of R_{Ω} , R_{SEI} and R_{CT} resistances, the total resistance is defined as

$$R_{\text{total}} = R_{\Omega} + R_{SEI} + R_{CT} \quad (20)$$

The total resistance computed above is plotted in Figure 19b. Similar to diffusion resistance, the total resistance is observed to also increase with the progression of days.

Figure 20 shows d-ohm metric computed using (12) for the battery that underwent three trials for three days. From the figure, it can be noticed that the ratio follows the same increasing pattern as that of the diffusion and total resistances shown in Figure 19.

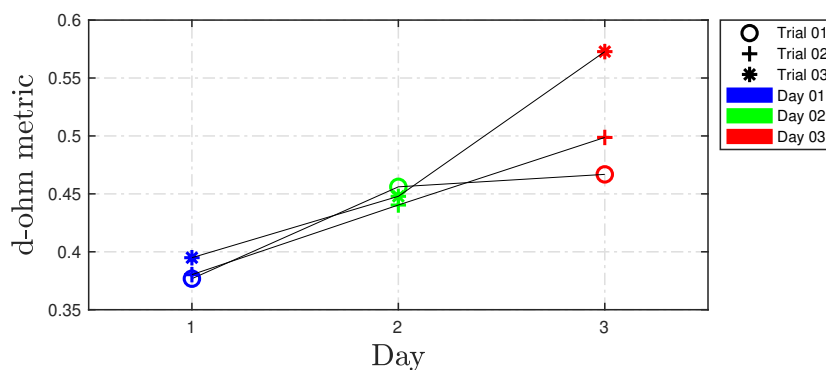


Figure 20. Ratio computed using (12).

4. Discussions

This paper investigates the uncertainty present in impedance spectra obtained from conventional Electrochemical Impedance Spectroscopy (EIS) measurements. For the first time, mathematical models are employed to characterize this uncertainty, and a novel d-ohm metric is introduced to quantify variations in both the impedance spectrum and the corresponding equivalent circuit model (ECM) parameters. The d-ohm metric is broadly applicable, being independent of battery chemistry, size, and experimental conditions. Its utility is demonstrated through an analysis of published datasets covering diverse chemistries, states of charge (SOC), and temperatures, as well as through repeated EIS measurements on a battery cell under controlled laboratory conditions.

Beyond the well-established factors influencing the impedance spectrum, such as state of health, SOC, temperature, and rest time, this work identifies drift as an additional source of variability. Repeated trials on the same cell at fixed SOC and temperature reveal measurable changes in the impedance spectrum, which manifest as bias and variance in ECM parameter estimation and can adversely affect SOH diagnostics. The findings highlight that current EIS-based diagnostic methodologies often overlook these systematic variations. By computing the mean impedance from multiple EIS measurements, the proposed approach mitigates drift-induced effects, providing a more robust and accurate framework for battery SOH estimation.

We identify several avenues for future research to further address variability in the impedance spectrum. Future works will aim to determine whether the identified drift effects persist under varying current profiles and to separate the respective contributions of temperature changes, ageing, rest time and other factors that lead to variations in the measured impedance. Further, systematic investigation is needed to assess whether the trends observed over three days in the battery continue beyond this timeframe and if/when the drift in the impedance spectrum stabilizes. All these future studies will help refine the d-ohm metric's applicability across all factors affecting the impedance spectrum and allow battery researchers to accurately predict the SOH of a battery.

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and K.P.; project administration, Bal.B.; All authors have read and agreed to the published version of the manuscript.

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